

**Networks Signals and Systems**  
**Prof. T. K. Basu**  
**Department of Electrical Engineering**  
**Indian Institute of Technology, Kharagpur**  
**Lecture - 29**  
**Tutorial**

We will continue with the numerical problem that we are discussing in the last class.

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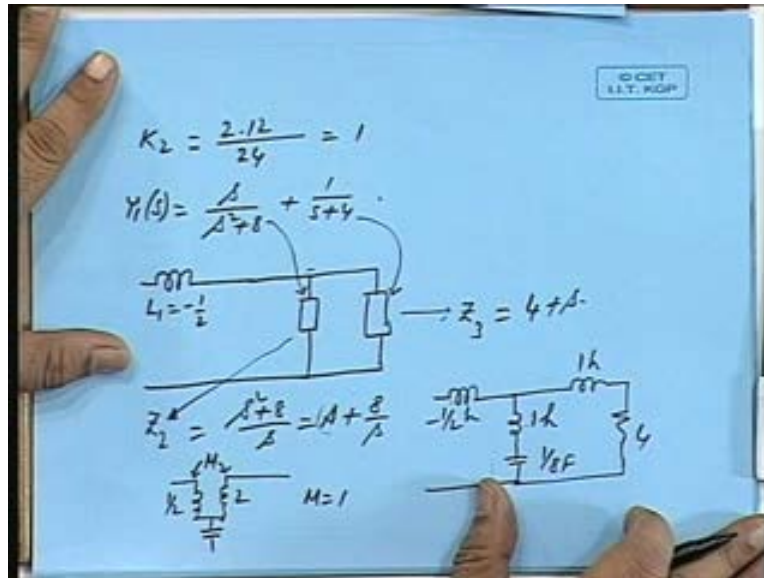
The image shows a handwritten derivation on a blue board. At the top right, there is a small logo that reads "© IIT KGP". The derivation starts with the function  $z(s) = \frac{s^2 + 2s + 16}{s^2 + 2s + 4}$ . It then identifies the poles of the denominator as  $\omega_1 = \sqrt{8}$  and  $\omega_2 = -\frac{1}{2} + j$ , with a corresponding real part  $\alpha = 4s$ . A circuit diagram shows a series combination of a resistor labeled  $-\frac{1}{2} \Omega$  and an inductor labeled  $L_1(s)$ . The equation  $z_1(s) = z(s) - L_1(s)$  is written, followed by the simplified expression  $\frac{(s^2 + 8)(s + 4)}{2(s^2 + 2s + 4)}$ . Next, the partial fraction decomposition is shown as  $Y_1(s) = \frac{2(s^2 + 2s + 4)}{(s^2 + 8)(s + 4)} = \frac{k_1 s}{s^2 + 8} + \frac{k_2}{s + 4}$ . Finally, the value of  $k_1$  is calculated as  $k_1 = \frac{2(-4 + 25)}{5^2 + 45} = \frac{-8 + 45}{-8 + 45} = 1$ .

We took a function  $z(s)$  equal to  $s$  squared plus  $2s$  plus  $16$  divided by  $s$  squared plus  $2s$  plus  $4$  we started off with this function and we found that at  $\omega_1$  is equal to  $\sqrt{8}$  the real part vanishes and we also obtained the value of  $x_1$  as half  $s$  rather corresponding impedance, this was  $L_1(s)$  all right and this was minus half  $s$  minus half Henry. This is  $z_1(s)$  which will be equal to  $z(s)$  minus  $L_1(s)$  and that give me  $L_1(s)$  is having a negative sign. So that give me a plus sign so we got  $s$  squared plus  $8$  into  $s$  plus  $4$  divided by  $2$  into  $s$  squared plus  $2s$  plus  $4$  okay. To realize a pole, to realize a pole we know we can make partial fractions, so this  $0$  is to be converted to a pole.

So we can very easily realize  $z_1$  is in terms of admittance function which will be  $2$  into  $s$  squared plus  $2s$  plus  $4$  divided by  $s$  squared plus  $8$  into  $s$  plus  $4$ , we write as  $k_1(s)$  by  $s$  squared plus  $8$  we know whenever the error routes on the imaginary axis it will be realized in terms of an LC network  $k_1(s)$  by  $s$  squared plus  $8$  plus  $k_2$  by  $s$  plus  $4$  this is to be seen later.

Let us see what this means how much is  $k_1$  multiplied by  $s$  squared by 8 divided by  $s$  squared plus 8 equal to 0. So this will give me 2 into 4 minus 8, so minus 4 plus 2  $s$  divided by  $s$  into  $s$  squared plus 8, so  $s$  into  $s$  plus 4, so  $s$  squared plus 4  $s$  which gives me this is minus 8 plus 4  $s$  divided by minus 8 plus 4  $s$ . So that is equal to 1.

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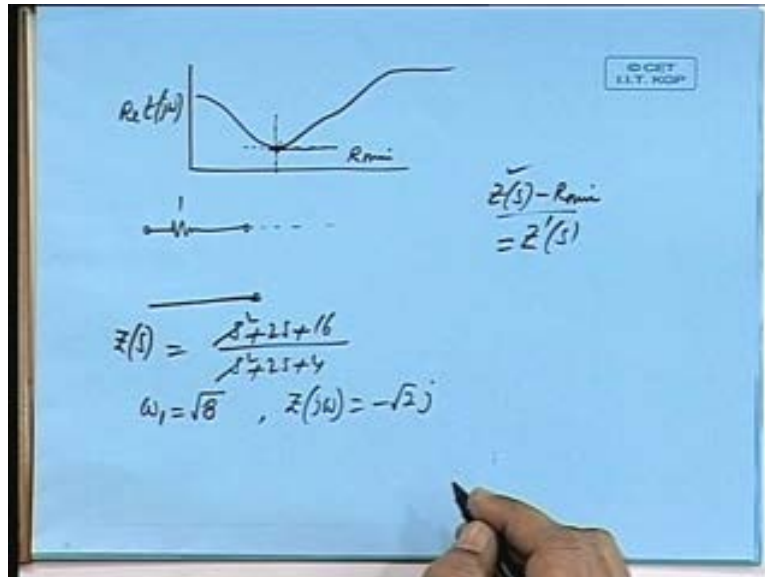


Similarly,  $k_2$  if I multiplied by  $s$  plus 4 and make  $s$  plus 4 equal to 0 say  $s$  is equal to minus 4, so 16 plus 4, 20 minus 2 into 48, so 12, so 2 into 12 divided by this is 16 plus 8, 24, is not, 24, so that is equal to 1. So  $y_1$  turns out to be  $s$  by  $s$  squared plus 8 plus 1 by  $s$  plus 4. So what is it mean you have got  $L_1$  equal to minus half Henry then we have got  $y_1(s)$  plus  $y_2(s)$  okay, this is  $y_2$ , this is  $y_1$ . So corresponding  $z_1$  or this  $z$  is  $s$  squared plus 8 by  $s$  which is 1 Henry plus 1, 8h farad and this one is, this  $z$ , if I call it  $z_2$  and  $z_3$  then  $z_3$  is 4 plus  $s$  all right.

So the network looks like this. This is 1 Henry, this is 18 farad, this is minus half Henry, this is 1 Henry and this is 4 ohms and this 1 Henry, 1 Henry and minus half Henry, this can be combined together in the form of a transformer 1 minus half. So this will be half Henry primary inductance, 2 Henry secondary inductance and 2 into half square root of that is 1, 1 Henry, m 100 percent coupling and then you have the capacitor, capacitor and resistor. So this is the network, **what does a network** sorry, Brune's network, is it all right?

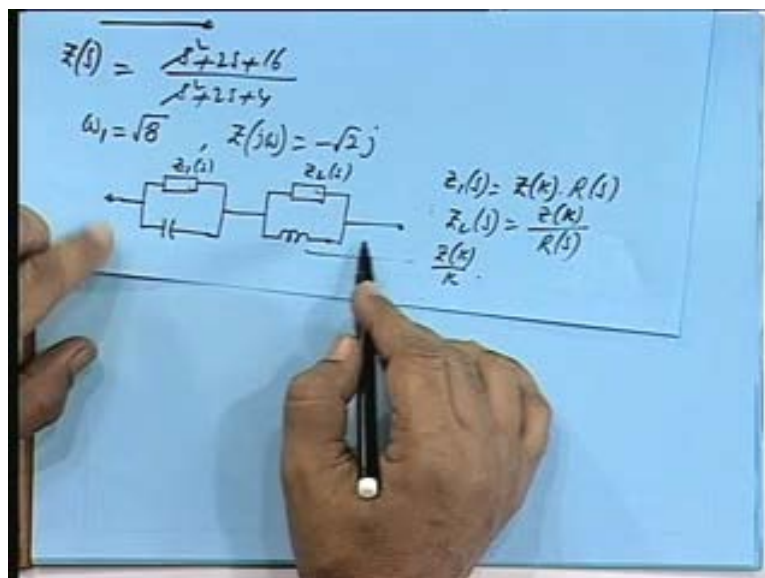
Let us see, what it will look like if we go for Bott Duffin synthesis, if you go for Bott Duffin synthesis, our starting point will be this omega 1 all right, whether that reactance where the frequency omega 1 has been identified corresponding to real part 0 okay any question at this point if the real part is not 0 suppose it is having a minimum value but it is not 0 it is minimum at certain value then we will take out that minimum value. Suppose we had given function where the real part varies like this.

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So we try to find out if it is not 0, if it is not vanishing at a particular frequency then we try to find out its minimum value then from the real part you separate out that. So we realize these are minimum separately okay and from this point onward our Brune's synthesis starts. so separate out suppose that comes out to be 1 ohm then you subtract from  $z(s)$  that 1 ohm whatever is a balance that will have a minimum point equal to 0 at this frequency, is not.

(Refer Slide Time: 10:55)



So this frequency has already been identified where this occurs and I have calculated the  $r$  minimum value separate it out subtract this  $r$  minimum from  $z(s)$  whatever is left over I call it  $z$  dash  $s$ , we will start the synthesis procedure from this point either by Bott Duffin or Brune's synthesis technique, is that point clear, because I want the real part should vanish at a particular frequency, so what is that frequency we have identified the frequency at which the value is minimum but it is not 0 say then if we extract out that resistance from the original  $z(s)$  then whatever is left over that will have a minimum equal to 0 at that frequency. So rest of the network now it becomes a minimum function earlier then function  $z(s)$  given may not be a minimum function all right. So we will take up this  $z(s)$  by Bott Duffin synthesis, so  $s^2 + 2s + 16$  divided by  $s^2 + 2s + 4$ , we found at  $\omega = 1$  is equal to  $\sqrt{8}$ ,  $j\omega$ ,  $j\omega$  was how much minus  $\sqrt{2}$ ,  $j$ , minus  $\sqrt{2}$   $j$  is that all right minus  $\sqrt{2}$   $j$  this will be our starting point.

Now what are the different functions that we took for Bott Duffin synthesis, the structure was a capacitor and then  $z_1(s)$  and an inductor and  $z_2(s)$  okay where,  $z_1(s)$ ,  $z_1(s)$  if you remember what did you write for  $z_1(s)$  yesterday,  $z_1(s)$  and  $z_2(s)$  that we wrote if you  $k$  times  $z(k)$  times  $R(s)$ , is it not and  $z_2(s)$  was  $z(k)$  by  $R(s)$  that reach at function  $L$  was  $z(k)$  by  $k$  all right and this was  $1$  by  $kz(k)$ . So if this is negative, this reactance is negative at that frequency, negative reactance comes only out of a capacitor. So at that frequency I want the current to should current should flow through this that means they this should be infinity and this should be a short if this is an infinite impedance this is a short because  $1$  is  $R(s)$  into something the other  $1$  is  $1$  by  $R(s)$  into that something all right.

So if this is infinity this will be 0 if this is infinity this will be 0 in this case what do you want if it is minus  $j_x$  then current should flow through this that means this is a reactance that should come into picture so it will be minus  $j_x$  plus 0, if it is plus  $j_x$  then this should be the path. So this should be short and current should flow like this this should be open is that all right. So you know when this will be infinity this will be 0 and this will be 0 this will be infinity if it is turning out to be plus  $j_x$  then it will be corresponding to  $j\omega L$ .

So this will be infinity this will be 0, so current will be short and then this reactance. So the reactance seen is this when it is minus  $j_x$  then reactance seen is this that means this will be open and this will be short so the overall impedance is this. So follow that procedure it is very simple so how much is  $C$ ,  $C_1$ , if I call it  $C$  okay or  $C_1$  whatever we call it, will be  $1$  by  $\omega = 1$  into  $z(c)$  which is  $1$  by  $\omega = 1$  is  $\sqrt{8}$ ,  $2\sqrt{2}$  okay  $\omega = 1$  is already identified and what was the magnitude of this reactance that was  $\sqrt{2}$ , is it not,  $z$  at that frequency is  $\sqrt{2}$  magnitude is  $\sqrt{2}$ . So this is the value of the capacitance, so that is equal to  $1$  by  $4$  farads. Now  $z_c(s)$  is nothing but  $kz(k)$  by  $s$  okay,  $z_c(s)$  is  $z(k)$  by  $k$  is  $L$  all right and  $C$  was  $1$  by  $kz(k)$ ,  $C$  is  $1$  by  $kz(k)$  and that has already been computed as  $1$  by  $4$ .

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$$C = \frac{1}{4z_c} \quad \omega_r = \sqrt{8} = 2\sqrt{2}$$

$$= \frac{1}{2\sqrt{2} \cdot \sqrt{2}}$$

$$= \frac{1}{4} F$$

$$z_c(s) = \frac{Kz(k)}{s} \quad C = \frac{1}{Kz(k)} = \frac{1}{4}$$

$$Kz(k) = 4$$

$$k \cdot \frac{k^2 + 2k + 16}{k^2 + 2k + 4} = 4$$

$$0 \Rightarrow (k-2)(k^2+8) = 0$$

$$k = 2$$

So  $kz(k)$  is 4 all right, so  $k$  into put  $z$  is equal to  $z(k)$  in place of  $s$  you put  $k$  so  $k$  squared plus twice  $k$  plus 16 divided by  $k$  squared plus twice  $k$  plus 4. So that is equal to 4 solve for this we get  $k$  cubed plus 2  $k$  squared minus 4, 4  $k$  squared so minus 2  $k$  squared okay plus 16  $k$  minus 8  $k$ , so plus 8  $k$  plus 16 comes to this side minus 16 equal to 0 or  $k$  minus 2 into  $k$  squared plus 8 equal to 0. So the real value of  $k$  is 2 okay, real value of  $k$  is 2.

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$$\therefore z(k) = \frac{4}{k} = \frac{4}{2} = 2$$

$$L = \frac{z(k)}{k} = 1$$

$$R(s) = \frac{kz(s) - sz(k)}{kz(k) - sz(s)}$$

$$= \frac{-2 \left( \frac{s^2 + 2s + 16}{s^2 + 2s + 4} \right) + 8 \cdot 2}{-4 + \frac{s^2 + 2s + 16}{s^2 + 2s + 4}}$$

$$= \frac{2(s^3 + s^2 + 2s - 16)}{s^3 - 2s^2 + 8s - 16} =$$

So how much is  $z(k)$ ,  $k$  into  $z(k)$  is  $4k$  is  $2$ , so  $z(k)$  is also  $2$ ,  $4$  by  $2$  is that okay therefore  $z(k)$  is  $4$  by  $k$  is equal to  $4$  by  $2$ . So we have got  $z(k)$  and  $k$  so how much will be  $L$ ,  $z(k)$  by  $k$  is it not,  $L$  is  $z(k)$  by  $k$ ,  $C$  is  $1$  by  $kz(k)$ ,  $z_2(s)$  is this  $z_1$  is this, so  $L$  is  $1$  okay  $L$  is  $1$ , what is function  $R(s)$  we have to compute now  $R(s) kz(s)$  minus  $sz(k)$  sorry it is drying up divided by  $kz(k)$  minus  $sz(s)$  substitute the value of  $k$ .

So it is  $2$  into  $s$  squared plus twice  $s$  plus  $16$  divided by  $s$  squared plus  $2s$  plus  $4$  minus  $s$  into  $z(k)$  is  $2$  divided by  $k$  into  $z(k)$  is  $4$  minus  $s$  into  $z(s)$ ,  $s$  squared plus  $2s$  plus  $16$  by  $s$  squared plus  $2s$  plus  $4$  okay. If you multiplied by  $s$  squared plus  $2s$  plus  $4$  and simplify then you get twice  $s$  cubed and change the sign we can put this as plus, this as plus, this as minus, this as minus and then twice  $s$  cubed. You can even  $2$  take  $2$  common, so  $2$  into  $s$  cube plus  $2$ ,  $2s$  are  $4$  minus  $2$ ,  $2s$  are  $4$ , thus check what you get my getting all right  $2$  into  $s$  squared plus  $z(s)$  just check the values I might have made a mistake  $2s$  are  $4$  and  $2s$  into  $2s$   $4$ , I made a slip somewhere  $s$  squared twice  $s$  squared sorry, this is  $2s$  squared, this is  $4s$  squared. So  $s$  cube plus  $s$  squared plus  $2s$  minus  $16$  all right, if we take  $2$  common so  $16$  divided by we will get  $s$  cubed minus  $2s$  squared plus  $8s$  minus  $16$  okay  $s$  cube plus  $2s$  squared minus  $4s$  squared so minus  $2s$  squared plus  $16$  okay.

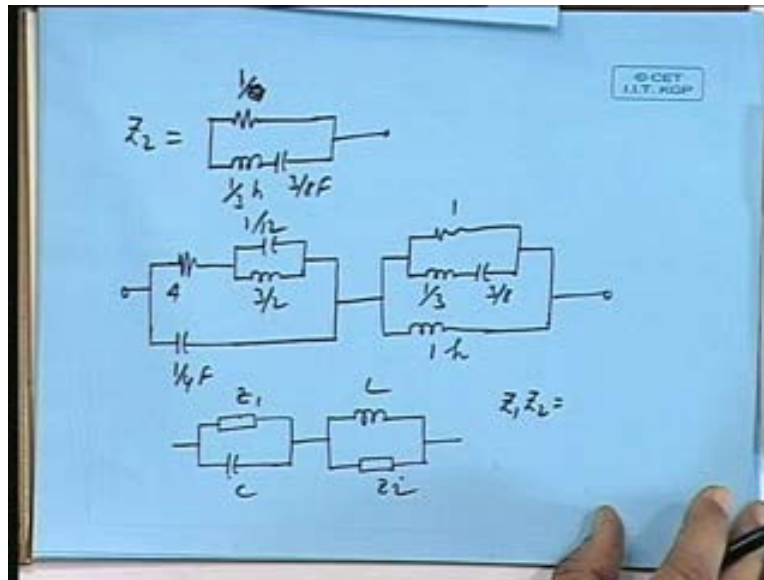
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The image shows a handwritten derivation on a blue board. At the top, a fraction is simplified:  $\frac{2(s-2)(s^2+3s+8)}{(s-2)(s^2+8)} = \frac{2(s^2+3s+8)}{s^2+8}$ . Below this,  $Z_1 = Z(k) \cdot R(s)$  is calculated as  $2 \cdot R(s) = 4 \frac{(s^2+3s+8)}{s^2+8}$ . This is further simplified to  $4 + \frac{12s}{s^2+8}$ . To the right,  $Z_2(s) = \frac{Z(k)}{R(s)} = \frac{2(s-2)}{2(s^2+3s+8)}$  is shown, which simplifies to  $Z_2 = \frac{s-2}{s^2+3s+8}$ . At the bottom left, a circuit diagram is drawn showing a voltage source of  $9$  V in series with a parallel combination of a  $3/2 \Omega$  resistor and a branch containing a capacitor  $C$  and an inductor  $L$  in series.

So I have got it will not come here equal to  $2$  into  $s$  minus  $2$  into  $s$  squared plus  $3s$  plus  $8$  divided by  $s$  minus  $2$  into  $s$  squared plus  $8$ . So  $s$  minus  $2$  will get cancelled okay you get  $2$  into  $s$  squared plus  $3s$  plus  $8$  divided by  $s$  squared plus  $8$  okay. So  $z_1$  which is  $z(k)$  into  $R(s)$  will be  $z(k)$  is  $2$ ,  $2$  into  $R(s)$  is equal to  $4$ ,  $2$  into this, so  $4$  into  $s$  squared plus  $3s$  plus  $8$  divided by  $s$  squared plus  $8$ . You can see for yourself  $s$  squared plus  $z$  by  $s$  squared plus  $z$  is  $1$ ,  $4$  plus  $12s$  by  $s$  squared plus  $8$  this is  $z_1$ .

So what is it 4 ohms and an LC parallel combination which is 12 Henry okay, 12 s by s squared 1 by 12 farad, 1 by 12 farad and 12 s by 8, 3 by 2 Henry, this is  $z_1$  okay. Similarly,  $z_2$  is  $z(k)$  by  $R(s)$  and that is equal to 2 by  $R(s)$  2 and 2 will go, go so this is s squared plus 8 divided by s squared plus 3 s plus 8 okay. So  $z_2$  I can realize as  $y_2$  which is s squared plus 3 s plus 8 by s squared plus 8 and which will give me same factors that will be in the form of an admittance.

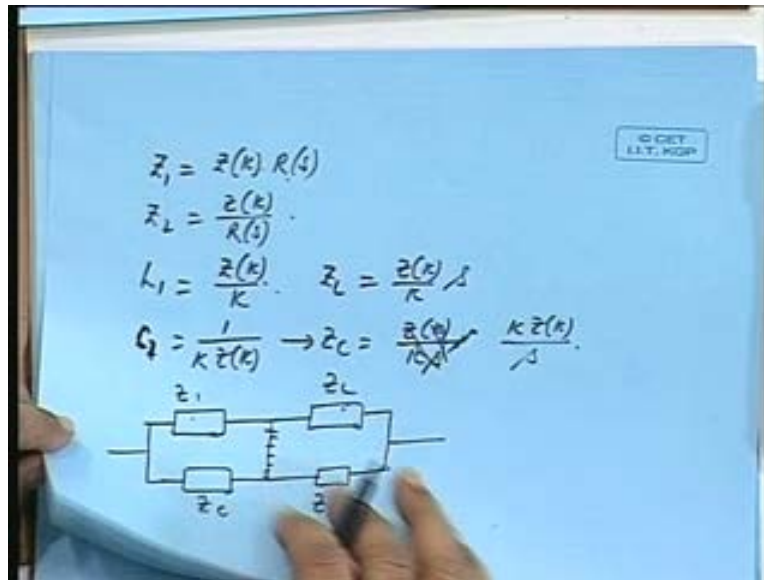
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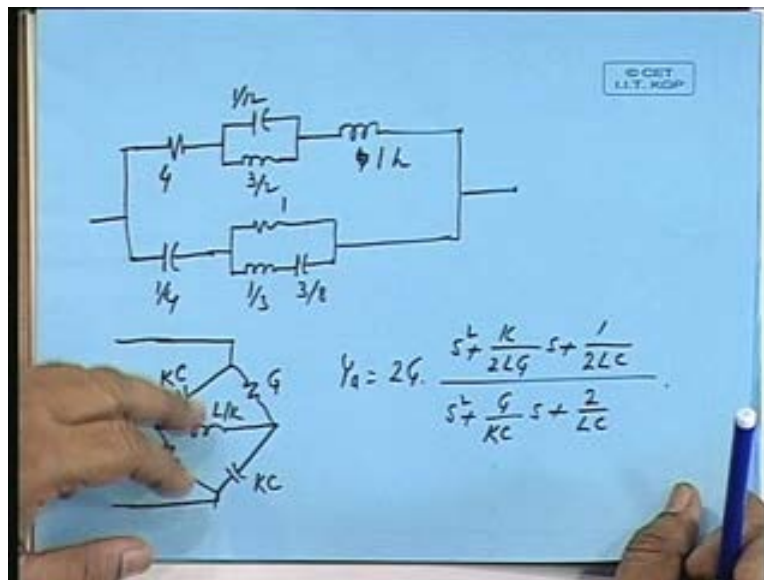
So 4 mho that  $z_2$  part will be therefore one fourth, this is nothing but this same factor sorry 1 plus sorry 1 plus 3 s by s squared plus 8. Now this 4 constant is not there so 1 mho and on this side we will have s squared plus 8 by 3 s, so s by 3, 1 third Henry and 8 by 3 s so 3 by 8 farad. So this is  $z_2$ , so finally you got a capacitor C that was how much was  $C_1$  by 4 farad and then with that we got  $z_1$  and  $z_1$  was 4 ohm and LC okay then L and how much was our  $L_1$  Henry and on this side  $z_2$  you have got this is a realization okay this is 4 this is 3 by 2 this is 1 by 12 farad, this is 1 ohm, this is 1 third and this is 3 by 8, so this is the Bott Duffin network one thing is very interesting. We have got c  $z_1$  this is shorted I might as well write this since it is a parallel combination L and  $z_2$  there is nothing wrong in it, is it not, is this all right.

Now you see this into this is equal to this into this, is it not how much is this  $z(k)$  by k is 1 by k into  $z(k)$ . So take the products how much was  $z_1$  and  $z_2$ ,  $z_1$  into  $z_2$  how much is it okay let us see sorry I will write all these  $z_1$  was  $z(k)$  into  $R(s)$   $z_2$  is  $z(k)$  by  $R(s)$ ,  $L_1$  how much was  $L_n$ ,  $L_1$ ,  $z(k)$  by k and  $L_2$  sorry C,  $C_1$ , how much is  $C_1$ , how much is  $C_1$ , no I am talking in terms of  $z(k)$ 's in terms of z's, in terms of z's we have written 1 is  $z(k)$  by  $R(s)$ , 1 by  $kz(k)$ , 1 by  $kz(k)$  is that all right. So  $z_1$ ,  $z(k)$  okay  $z_1$  into  $z_2$  is  $z(k)$  squared  $z_1$  into sorry  $z_1$  into L how much is it  $z(k)$  squared into  $R(s)$  by k, what is  $z_2$  into this is  $C_1$  sorry. So how much is this  $z_L$ ,  $z(k)$  by k into s, how much is  $z_c$ ,  $z_c$  is  $z(k)$  by k into s, is it not sorry 1 by C was 1 by  $kz(k)$ , is it so 1 by cs, so  $kz(k)$  by s.

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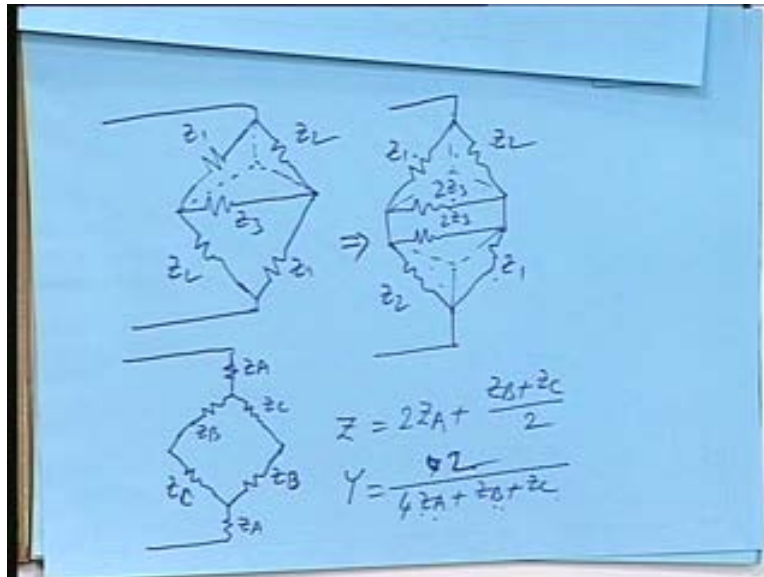
Now how much is  $z_1$  into  $z_2$ ,  $z_k$  squared, how much is  $z_L$  into  $z_c$  also  $z_2$  squared. So if I write  $z_1$ ,  $z_c$ ,  $z_L$  and  $z_2$  okay this is what we have written the same thing I am writing in in place of L and C I am just writing  $z_L$  and  $z_c$  take the product of these 2 is equal to this. So it is a balanced bridge so this can be eliminated this into this is equal to this into this. So it is a balanced bridge network, so this can be removed so I can put  $z_1$  in series  $z_L$ ,  $z_c$  in series  $z_2$ , impedance function is same is it not. So this same network one fourth farad in series with this and 1 Henry inductor in series



with this both of them will be in parallel. So this network will be therefore reduced to 4 ohms, 3 by 2, 1 by 12, 1 Henry. This is 4 ohms 3 by 2, 1 by 12, 4 Henry, 1 Henry and this side will have 1 by 4, 1 ohm, 1 third and 3 by 8. So this is the overall realization you could have left at this point also is that 2 series drops here there are 2 parallel blocks okay.

It is an interesting question that has been asked here most of problems are taken from some of the very standard books you may come across in many places okay this interesting question is determine the admittance okay I will just draw this first. This is given as an admittance  $G$ , this is given  $kc$ , this is given as  $L$  by  $k$ , this is given as  $G$ , this is given as  $kc$ , the admittance seen from this side  $y_a$  show that  $y_a$  is twice  $G$  into  $s$  squared plus  $k$  by  $2 LG$ ,  $s$  plus  $1$  by  $2 LC$  divided by  $s$  squared plus  $G$  by  $kc(s)$  plus  $2$  by  $LC$ . Obviously, the admittance is appearing to be quiet combustion, is it not one may go for start to delta conversion then again we will get parallel branches okay, we can do it. I thought that I go for a little short cut I mean it is easy to compute.

(Refer Slide Time: 33:45)



Okay it is not otherwise difficult you are you have been given say I will write in the very general form  $z_1$ , something like this  $z_2, z_3, z_2, z_1$ . So if I reduce this I am start to delta and then again add with this this add with the  $z_2$  make the parallel combinations and then add with this series okay that because of this symmetric positions  $z_1$  and  $z_2$ , suppose  $z_3$  we replace by twice  $z_3$  then I will have a delta here twice  $z_3$  in parallel with that another twice  $z_3$  okay.

Now this one will give me a star product okay, let me show it like this say  $z_a, z_b$  and  $z_c$  then what will be the products of this  $z_1, z_2$  twice  $z_3$  this one, this also  $z_1, z_2$  twice  $z_3$  it will be  $z_a$  I am sorry  $z_c, z_b$  just  $z_b$  and  $z_c$  will interchange their positions and this will remain as  $z_a$ , is it not. See this one and this one they are identical except that  $z_1, z_2$  I have just rotated this, is it not. So the same element values will come in just change the order, so  $z_b$  and  $z_c$  will be interchanging their positions.

Now it is  $z_b$  plus  $z_c$  on this side this side also  $z_b$  plus  $z_c$  so parallel combination of these 2, so what will be the total  $z$  will be  $z_a$  plus  $z_a$  twice  $z_a$  plus  $z_b$  plus  $z_c$  divided by 2, is that all right. So make a tabulated conversion of these impedances  $z_1$ ,  $z_2$  and twice  $z_3$  okay and then compute this. So this is quite a bit of reduction in level in the computational level once we know the  $z$  we can calculate  $y_1$  over this, so this is 4 by 4  $z_a$  plus  $z_b$  plus  $z_c$ . So I leave it as an exercise you can do this and derive this relation very easily there is no problem yes, and sorry this is 2, 2 by when you put the values of  $z_a$ ,  $z_b$ ,  $z_c$  you will get that in this form. Now that we have got the method of rlc synthesis ready you remember sometime back we checked  $s$  plus 2 into  $s$  plus 3, is it into this plus 4, is it a positive real function and if it is realizable can you realize one. So  $m_1$ ,  $m_2$  minus  $n_1$ ,  $n_2$  will check that time which is  $s$  squared plus 5  $s$  plus 6 divided by  $s$  squared plus 5  $s$  plus 4.

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The image shows a handwritten derivation on a blue board. At the top right, there is a small logo that reads "G-CET I.I.T. KGP". The derivation starts with the equation:

$$\frac{(s+2)(s+3)}{(s+1)(s+4)} = \frac{s^2 + 5s + 6}{s^2 + 5s + 4}$$

Below this, the numerator minus the denominator is calculated:

$$m_1 m_2 - n_1 n_2 = (s^2 + 6) - 25s = s^2 - 25s + 6$$

$$= s^2 + 15s + 24$$

Then, the partial fraction expansion is shown:

$$\Rightarrow \frac{2}{3(s+1)} + \frac{2s}{12(s+4)} + \frac{5}{6}$$

Next to this, the values of the constants are given:

$$k_2 = \frac{1}{6}$$

$$1 = \frac{1}{3} + k_3$$

At the bottom, a circuit diagram is drawn. It consists of a parallel combination of a resistor with value  $\frac{2}{3}$  and a branch containing a resistor with value  $\frac{2s}{12}$  in series with a capacitor with value  $\frac{1}{6}$ . The output terminals are labeled with a voltage of  $\frac{5}{6}$ .

So  $s$  squared plus 6 into  $s$  squared plus 4 minus 25  $s$  squared, so that give me  $s$  to the power 4 minus  $s$  squared plus minus 15  $s$  squared 6 plus 4, 10 minus 25, 15  $s$  squared plus 24, so if I put  $s$  equal to  $j$  omega that gives me omega to the power 4 plus 15 omega squared plus 24, this is always positive. So if I am asked to realize this function wholes and 0s are not coming alternately, so this is going to be an rlc combination, rlc combination, okay.

So one may try partial fractions. Let us try if it too if it works otherwise we will go for that Bott Duffin or Brune's synthesis, if it is possible to write this as  $k_1$  by  $s$  plus 1 plus  $k_2$  by  $s$  plus 4 or it can be  $k_1(s)$  by  $s$  plus 1 plus sorry  $k_2$  by  $s$  plus 4 that means I put  $s$  with one of them one is RC the other one is RL plus if required a constant  $k_3$ , either of them. Let us see whether we can get in this form in the first one if I write in this form how much is  $k_1$  put  $s$  plus 1 is equal to 0. So that gives you 1 into 2 by 3 so 2 by 3  $s$  plus 1 okay  $k_2$  if I multiply by  $s$  plus 4 divide by  $s$  and then make  $s$  plus 4 equal to 0 then it will be 2 minus 2 minus 1 divided by minus 4 minus 3.

So it will be positive so this will be 2 by 4 into 3, 12, s by s plus 4 it is 1 by 6 plus  $k_3$  if I make  $s_{10}$  into infinity you get anything  $s_{10}$  into infinity will be  $k_2$  plus  $k_3$  okay  $k_2$  is how much 1 by 6 if I make  $s_{10}$  into infinity this will become  $s_2$  the part 2 by  $s_2$  the part 2 so 1, 1 on the right hand side if I make  $s_{10}$  into infinity it will be  $k_2$  plus  $k_3$ ,  $k_2$  is 1 by 6 plus  $k_3$ , so  $k_3$  is 5 by 6 okay. So this will be an RC network all right. 2 by 3 s so 3 by 2 farads  $s_{10}$  into 0 means 2 by 3 ohms resistance this is an RL network  $s_{10}$  into infinity means 2 by 12, 1 by 6 ohm resistance when extending to 0 it is 1 by 6 sorry when extend into infinity yes, s tending into 0 it will be 2 by 48 so 1 by 24 farads okay and then 5 by 6 ohm resistance. It is possible to realize simply by partial fractions only thing we will have to keep on trying possibilities either this or the other one is like this. It is an rlc you can also try by Brune's synthesis.

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Handwritten mathematical derivation on a blue background:

$$z(s) = \frac{s^2 + 5s + 6}{s^2 + 5s + 4} \quad (s^2 + 5s + 4) = 25s^2$$

$$Re\{z(j\omega)\} = \frac{m_1\omega - n_1\omega}{\omega^2 - n_2} = \frac{\omega^4 + 15\omega^2 + 24}{\omega^4 + 17\omega^2 + 16}$$

$$\omega = x \Rightarrow \frac{x^2 + 15x + 24}{x^2 + 17x + 16} = f(x)$$

$$f'(x) = 0 \Rightarrow (2x + 15)(x^2 + 17x + 16) = (2x + 17)(x^2 + 15x + 24)$$

$$\Rightarrow 17x^2 + (22 + 255)x + 240 = 42x^2 + (48 + 255)x + 17 \cdot 24$$

$$\Rightarrow 22x^2 + 11x + 168 = 0 \quad x = \frac{-4 \pm \sqrt{11}}{7}$$

$$7x^2 + 42x + 42 = 0$$

Let us see when is that real part vanishing, does the real part vanish if not what to do? So  $z(s)$  equal to  $s$  squared plus  $5s$  plus  $6$  I am just rewriting this problem  $s$  squared plus  $5s$  plus  $4$ . So real  $z(s)$  real  $z(j\omega)$  will be  $m_1, m_2$  minus  $n_1, n_2$  by  $m_2$  squared minus  $n_2$  squared and that is equal to we just now worked it out  $s$  to the power 4 sorry  $\omega$  to the power 4 plus  $15\omega$  squared plus  $24$  divided by how much is  $m_2$  squared minus  $n_2$  squared a squared plus  $4$  whole squared, a squared plus  $4$  whole squared minus  $25s$  squared, put  $s$  equal to  $j\omega$ .

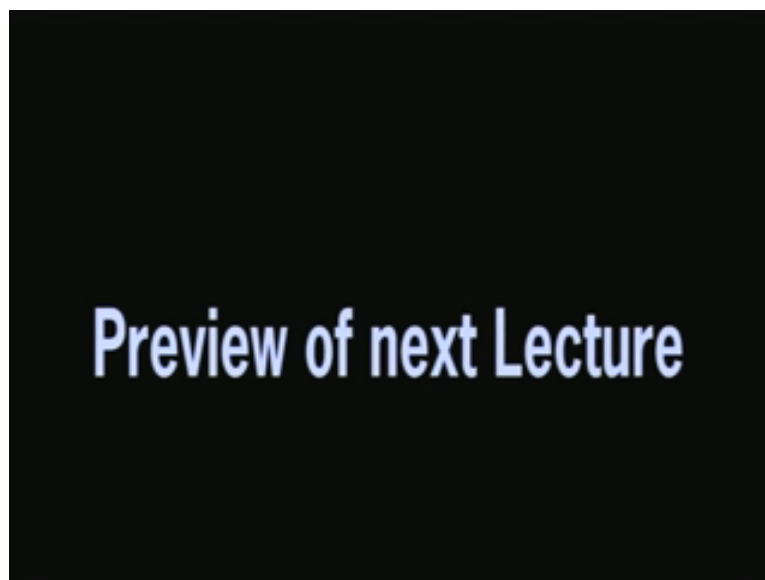
So  $\omega$  to the power 4 plus  $4$  into  $28$  minus  $25$  okay,  $17$  minus  $17\omega$  squared means plus  $17\omega$  squared, minus  $17s$  squared plus  $16$  all right. Now this is not vanishing at any frequency then what we do? what we do? Find out the frequency at which these value is minimum I can put  $\omega$  squared is equal to  $x$ , let me make it in a quantity form. So this will become  $x$  squared plus  $15x$  plus  $24$  divided by  $x$  squared plus  $17x$  plus  $16$  then let us call it  $f(x)$ . So  $f'(x) = 0$ , so that gives me how much twice  $x$  plus  $15$  into  $x$  squared plus  $17x$  plus  $16$  that will be equal to after equate to  $0$  twice  $x$  plus  $17$  into  $x$  squared plus  $15x$  plus  $24$ , correct if I am wrong, is that all right, twice sorry twice  $x$  cube will get cancelled then  $17x$

square  $34x^2 - 15$ ,  $34x - 15$ ,  $19x^2$  okay plus 16 into 2,  $32x + 17$  into 15 all right. So  $255x + 15$  into 16, 240 on this side we get  $2x^2 - 2x$  cube get cancelled then  $15x + 17$ ,  $47x^2$  all right plus 24 into 2,  $48x + 48x + 255x + 17$  into 24, is it all right  $17x + 24$ . So that gives me  $47x - 19$ ,  $28x^2 - 47x - 19$   $255x$  will get cancelled  $48x - 32$ ,  $16x + 17$  into  $24x - 24$  into 10, so  $7x + 24$ , is that all right,  $168$  if you divide by  $47x^2 + 4x + 42$  is equal to 0 is that all right  $7x^2 + 42x + 4x$ , so you will get one real value, positive value of  $x$ .

So  $x$  is let us calculate  $-4 \pm \sqrt{16}$ , we get imaginary values, **both are imaginary**, both are imaginary there is something something wrong I might have made a mistake that must be some frequency where it is minimum  $\omega^2 = x^2 + 15x + 24$  plus  $17x + 16$ . I hope this is all right okay then  $2x + 15$  into  $x^2 + 17x + 16$ ,  $2x + 17$  into  $x^2 + 15x + 24$ , so  $2x^2 - 2x$  cube they get cancelled  $2x$  into okay.

Let me rewrite it, anyway what I want it to state is you will get a real value of  $x$  that is equal to  $\omega^2$  so calculate  $\omega$  and substitute that  $\omega$  here in the real part. So that will give you the minimum value that is when the real part, real part varies like this, it is this minimum value and after computing that  $R$  minimum subtract it from  $z(s)$ , subtract it from  $z(s)$  whatever is left over you start realizing that  $z(s)$ , the remainder  $z(s)$ . There may be a small slip somewhere here we will discuss it in the next class, if time permits otherwise you work it out yourself and since there is not much of time okay, thank you very much. We will continue with this in the next class.

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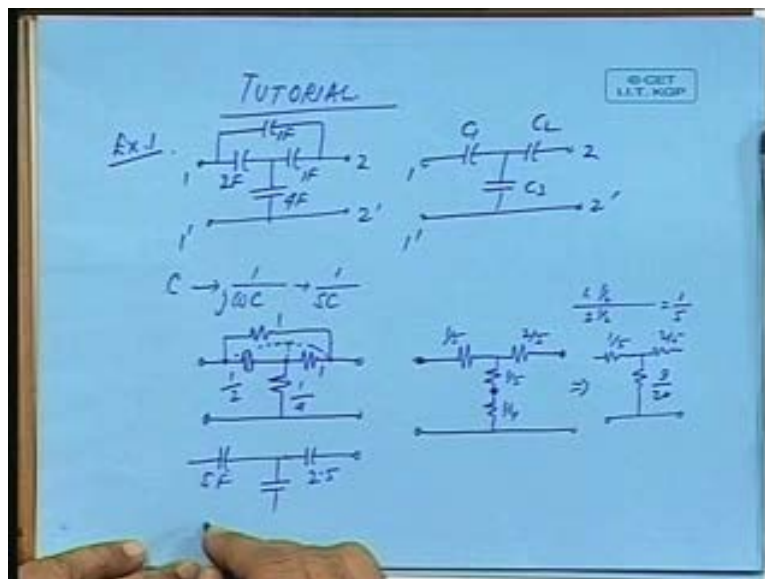
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# Lecture #30

# Tutorial

Good morning friends, today we will have a tutorial exercise on some of the topics that we have covered so far.

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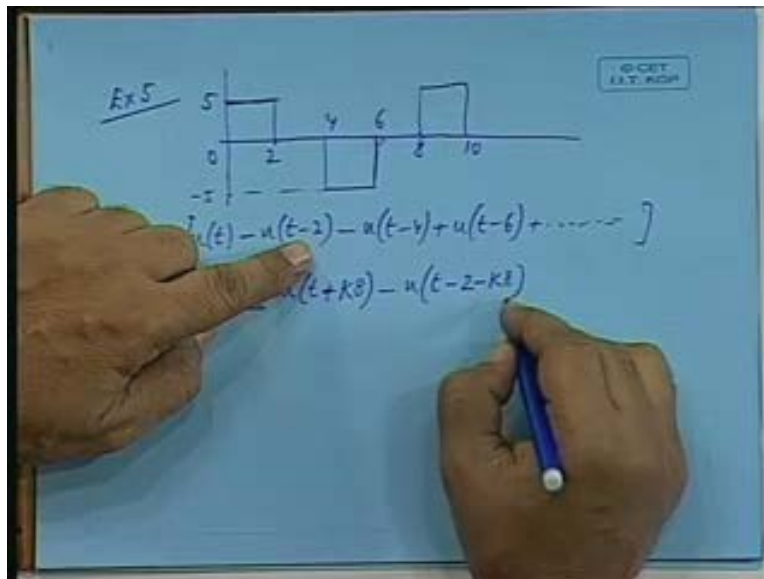
I will take up a few examples and discuss there is a problem, first example is a very simple one, you have bridged capacity circuit the values are 2 farads, 1 farad, 4 farads and 1 farad. What will be the equivalent capacitances  $C_1$ ,  $C_2$  and  $C_3$  okay. Now in this case as you know a capacitor is

having an impedance if you talk in terms of sinusoids, it will be  $1/\omega C$ , it is  $j$  here if you talk in Laplace domain it will be  $1/sC$  that means the impedance is inversely proportional to  $C$ .

So I can take these capacitances to be equivalent impedances or even resistances with the values which will be just inverse of this something like half ohm, one fourth ohm, 1 ohm and 1 ohm. I can replace it by an equivalent impedance circuit. Now after this we can get equivalent  $1/\omega C$  by a star delta conversion see if this is a delta then I can have a star like this. So these 3 nodes are these 3, so what will be the value of this  $1/\omega C$  into half divided by  $1 + 1/2 + 1/2$ . So  $1/\omega C$  into half divided by 2 and a half, so that gives me  $1/\omega C$  by 5,  $1/\omega C$  by 5.

Similarly, this one will be  $1/\omega C$  into  $1/2 + 1/2$ , so  $2/\omega C$  by 5 and similarly this is  $1/\omega C$  into  $1 + 1/2$ , so this one will be  $1/\omega C$  by 5, this is  $1/\omega C$  by 4, so you get  $1/\omega C$  by 5,  $2/\omega C$  by 5 and  $1/\omega C$  by 5 plus  $1/\omega C$  by 4,  $9/\omega C$  by 20. So take the inverse of these that will give you the equivalent capacitance 5 farads, 2.5 farads and 20 by 9 farads okay. So these are the 3 equivalent capacitances, what will be the expression for this type of periodic function? It is pretty simple 0, this is 5, 2, 4, 6, 8, 10 and so on, this is minus 5.

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Obviously the first block can be written as 5 into  $u(t)$  and then at 2 seconds I apply a negative step at  $u(t)$  minus 2. So that gives me the first block again at 4 it is negative so minus  $u(t)$  minus 4 and at 6 I apply  $u(t)$  minus 6 a positive step and that completes 1 period up to 8 seconds and then it keeps on repeating. So show 1 period if I show 1 period and then show it as a repetition of the same function this will be written as 5, I can write this as summation  $u(t)$  minus 2 okay  $u(t)$  so plus after 8 seconds again it is plus  $u(t)$  okay.

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Ex. 10

$Z(s) = \frac{s}{s^2+2s}$

$\frac{1}{s^2+2s} = \frac{a}{s} + \frac{b}{s+2}$

$(s+2)a + 2s = as^2 + (2a+2b)s = 5s^2 + 5bs$

$a = 5$   
 $2a + 2b = 5$   
 $10 + 2b = 5$   
 $2b = -5$   
 $b = -\frac{5}{2}$

$C = 2$

So  $u(t) + k_8 \sin u(t) - 2 \sin k_8$ ,  $C$  has to be equal to 2 okay,  $C$  has to be equal to 2. If I add this  $s^2 + 2$  into  $as + 2s$  that gives me  $s^3 + 2s$  and that is equal to  $5s^3 + 5s$  plus  $5bs$ . So here  $a$  is equal to 5 and if I put equal to 5, 5 into 2, 10 plus 2, 12 is equal to 5  $b$ , therefore  $b$  is equal to 12 by 5 okay, this is a very simple example. Well, before we take up any other problem I think we will stop here today, we will continue with the next class because there is not much of time.