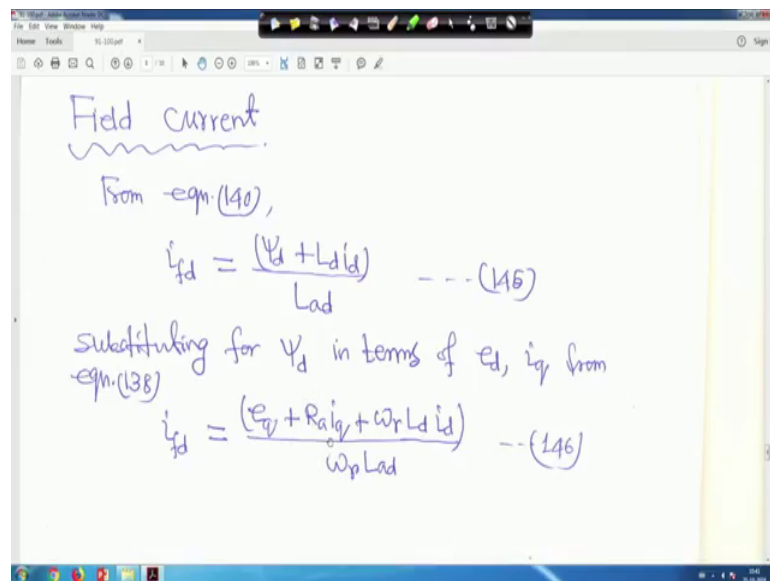


Power System Dynamics, Control and Monitoring
Prof. Debapriya Das
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Lecture - 10
Power System Stability (Contd.)

So, this is our i_{fd} is equal to this expression, right.

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Field current

From eqn. (140),

$$i_{fd} = \frac{\psi_d + L_d i_d}{L_{ad}} \quad \text{--- (145)}$$

Substituting for ψ_d in terms of e_d , i_q from eqn. (138)

$$i_{fd} = \frac{(e_q + R_a i_q + \omega_r L_d i_d)}{\omega_p L_{ad}} \quad \text{--- (146)}$$

Now, this $R_a i_q$ is the voltage drop and $\omega_r L_d i_d$, right. So, this we $\omega_r L_d i_d$, so generally $L \omega$ is reactance, so this form we can make $X_d i_d$, right. So, that is why this i_{fd} this term equation can be written as e_q plus $R_a i_q$ plus $X_d i_d$ and $\omega_r L_{ad}$ $L \omega$ generally reactance, so $\omega_r L_{ad}$, so this is X_{ad} . This is equation 147, right.

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Replacing the product of synchronous speed and inductance L by the corresponding reactance X

$$i_d = \frac{(E_g + R_a i_q + X_d i_d)}{X_d} \quad \dots (147)$$

The inductances/reactances appearing in eqn. (137) to (147) are saturated values.

Phasor Representation

Now, the inductances or reactances appearing in equation 137 to 147 they are basically we are using saturated values, right.

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Phasor Representation

For balanced steady-state operation, the stator phase voltages may be written as:

$$e_a = E_m \cos(\omega_s t + \alpha) \quad \dots (148)$$
$$e_b = E_m \cos(\omega_s t - \frac{2\pi}{3} + \alpha) \quad \dots (149)$$
$$e_c = E_m \cos(\omega_s t + \frac{2\pi}{3} + \alpha) \quad \dots (150)$$

So, next is the phasor representation. For balance steady state operation the stator phase voltages may be written as, right that is e_a is equal to $E_m \cos \omega_s t + \alpha$, right. And similarly e_b is equal to we can write $m \cos \omega_s t - 2\pi/3 + \alpha$ 120 degree phase shift. Similarly e_c is equal to $e_m \cos \omega_s t + 2\pi/3 + \alpha$. This is equation 150, right.

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$$e_c = E_m \cos(\omega_s t + \alpha) \quad (10)$$

where ω_s is the angular frequency and α is the phase angle of e_a with respect to the time origin. Applying the dq transformation gives

$$e_d = E_m \cos(\omega_s t + \alpha - \theta) \quad \dots (151)$$
$$e_q = E_m \sin(\omega_s t + \alpha - \theta) \quad \dots (152)$$

Now, where ω_s is the angular frequency and α is the phase angle of e_a with respect to the time origin. Now, applying the dq transformation for this thing earlier we have seen dq transformation if we apply the dq transformation it will give e_d will give $E_m \cos \omega_s t + \alpha - \theta$, this is equation 151. And e_q will give $E_m \sin \omega_s t + \alpha - \theta$, this is equation 152, right.

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The angle θ by which the d-axis leads the axis of phase-a is given by

$$\theta = \omega_r t + \theta_0 \quad \dots (153)$$

where θ_0 is the value of θ at $t=0$.

With ω_r equal to ω_s at synchronous speed, substitution for θ in eqns. (151) and (152) yields

Therefore, now the angle θ we have seen that figure 9, we have seen. The angle θ by which that d-axis lead the axis of phase a is given by you can write θ is equal to

$\omega_r t + \theta_0$ this is equation 153 and θ_0 is the value of θ at t is equal to 0, right.

Now, with ω_r equal to ω_s , right because it is run machine is running at synchronous speed say at synchronous speed substitution for θ in equation 151 and 152 it will give you e_d is equal to $E_m \cos \alpha - \theta_0$ that means, your this thing, your this one your here.

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The phase angle of I_a with respect to the time origin. Applying the dq transformation gives

$$E_d = E_m \cos(\omega_s t + \alpha - \theta) \dots (151)$$

$$E_q = E_m \sin(\omega_s t + \alpha - \theta) \dots (152)$$

$$\omega_s t + \alpha - (\omega_r t + \theta_0)$$

$$= \omega_s t + \alpha - \omega_r t - \theta_0$$

$$\omega_s = \omega_r = \omega$$

$$= \alpha - \theta_0$$

The angle θ by which the d-axis leads


This only we are writing say only one term say this is $\omega_s t + \alpha - \theta$ and θ is equal to $\omega_r t + \theta_0$, right. So, this one actually $\omega_s t + \alpha - \omega_r t - \theta_0$, but ω_r is equal to ω_s , right therefore, this term this term will be cancel. So, ultimately it will be say $\alpha - \theta_0$ this term will be $\alpha - \theta_0$, right.

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With ω_p equal to ω_s at synchronous speed, substitution for θ in eqns (151) and (152) yields

$$e_d = E_m \cos(\alpha - \theta_0) \quad \text{--- (154)}$$
$$e_q = E_m \sin(\alpha - \theta_0) \quad \text{--- (155)}$$


In the above equations, E_m is the peak value of phase voltage. In steady-state analysis



So, that is why here we are writing that e_d is equal to $e_m \cos \alpha$ minus θ_0 this is equation 154. And e_q is equal to $e_m \sin \alpha$ minus θ_0 that is equation 155, right. So, in the above equation that is E_m is the peak value of phase voltage, right.

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In the above equations, E_m is the peak value of phase voltage. In steady-state analysis we are interested in RMS values and phase displacements rather than instantaneous or peak values. Using E_t to denote per unit RMS value of armature terminal voltage and noting that in per unit RMS and peak values are equal

$$e_d = E_t \cos(\alpha - \theta_0) \quad \text{--- (156)}$$
$$e_q = E_t \sin(\alpha - \theta_0) \quad \text{--- (157)}$$


And in steady state analysis we are interested in RMS values and phase displacement rather than instantaneous or peak values. Now, using E_t to denote per unit RMS values of armature terminal voltage and noting that per unit RMS and peak values are equal because when we converting it to the per unit values RMS value and peak value they will

be equal. Therefore, we can write like this E_t is equal to $E_t \cos(\alpha - \theta_0)$ this is equation 156 and e_q will be $E_t \sin(\alpha - \theta_0)$ this is equation 157, right.

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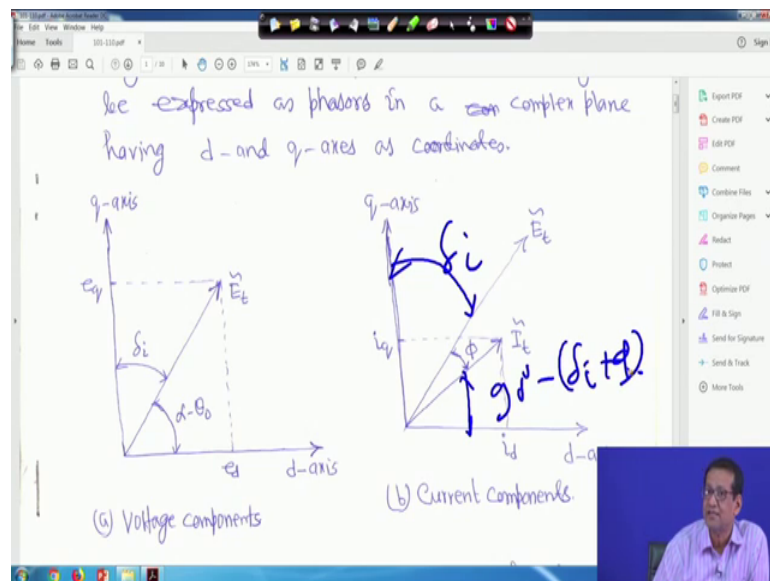
Now, just hold on now that dq components of armature voltage are scalar quantities this we have seen before. However, in view of the trigonometric relationship between them they can be expressed as phasors in a complex plane having d and q-axis as coordinates, right.

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For example, this is my, your what you call this is my q-axis; this is my q-axis and this is my d-axis. So, E_t is the phasor quantity that, right and this angle is δ_i and this angle your α minus your what you call θ_0 , E_t is equal to $E_t \cos(\alpha - \theta_0)$ and e_q is equal to $E_t \sin(\alpha - \theta_0)$ and this angle is δ_i .

Similarly, here also q-axis I have not written here this angle actually δ_i , right. So, similarly if you and this is say E_t , right and this E_t is the current. Therefore, E_t is lagging from your E_t , I_t lagging from E_t by an angle ϕ , right. So, in this case for this case your what you call that e_d is equal to your in general $E_t \cos(\alpha - \theta_0)$ whatever, we have seen before. Similarly, e_q is equal to $E_t \sin(\alpha - \theta_0)$ this side, right, but this angle δ_i will come later, right and this angle is δ_i , similarly for the current, right.

(Refer Slide Time: 06:17)



And this angle I told you this angle is δ_i . So, similarly for if you just see this is my δ_i this is δ_i and this is ϕ , right. So, this angle actually $90^\circ - \delta_i + \phi$, right. Therefore, i_d will be is equal to $I_t \cos(90^\circ - \delta_i + \phi)$ that is i_d is equal to $I_t \sin(\delta_i + \phi)$.

Similarly, i_q will be your $I_t \cos(\delta_i + \phi)$, right. So, this is voltage component and this is current components, right. And this is figure something marked 15 representation of dq components of armature voltage and current as phasor.

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Fig.15: Representation of dq components of armature voltage and current as phasors.

The armature terminal voltage may be expressed in complex form as

$$\tilde{E}_t = e_d + j e_q \quad \dots (158)$$

By defining denoting δ_i as the angle by which the q-axis leads the phasor \tilde{E}_t , equations (156) and (157) become

(Note: A small video inset of a man speaking is visible in the bottom right corner of the slide.)

Now, the armature terminal voltage maybe now it can be written in complex form that e_t will be $e_d + j e_q$. This is equation 158 because here you have this is e_d this is e_q . So, e_t we can write as $E_t \cos \delta_i + j E_t \sin \delta_i$. So, we are writing $e_d + j e_q$ this is 158.

Now, by denoting δ_i as the angle by which the q-axis lead the phasor E_t , right. Equation 156 and 157 it become that e_d is equal to $E_t \cos \delta_i$ and e_q is equal to $E_t \sin \delta_i$.

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$e_d = E_t \cos \delta_i \quad \dots (159)$

$e_q = E_t \sin \delta_i \quad \dots (160)$

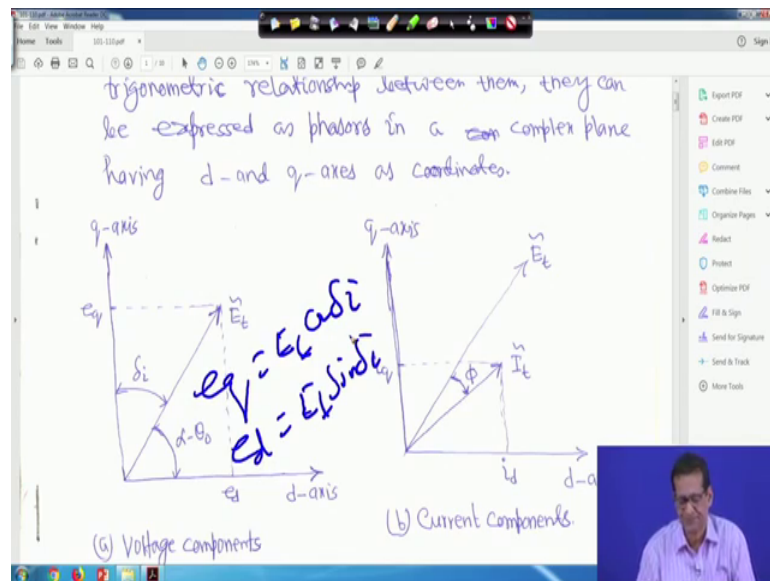
Similarly, the dq components of armature terminal current I_t can be expressed as phasors. If ϕ is the power factor angle, we can write

(Note: A small video inset of a man speaking is visible in the bottom right corner of the slide.)

Now, if you look into this; if you look into this then your a once again that if you look

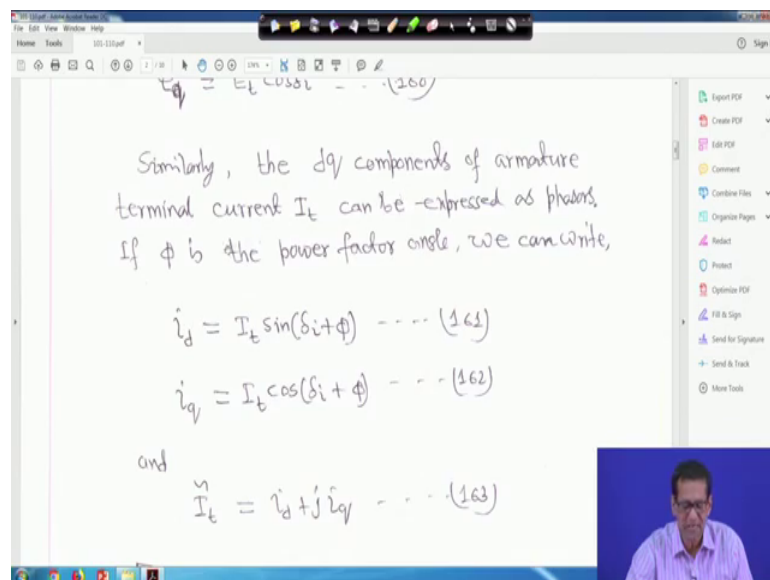
into this your e_q will be is equal to $E_t \cos \delta$ if you take δ i.

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And similarly e_d will be is equal to $E_t \sin \delta$, right so that is what we are writing here.

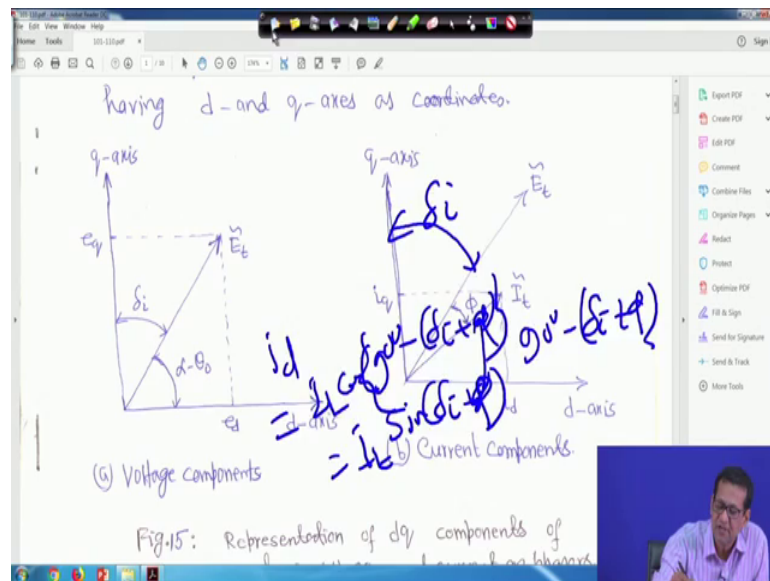
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So, similarly for i_d I told you similarly the dq component of the armature terminal current I_t can be expressed as phasor if ϕ is the power factor angle we can write i_d is equal to $I_t \sin \delta$ this I told you that this angle it is not marked here, but this angle same thing between q-axis and E_t , so this angle is δ , right and therefore, this angle

is 90 degree minus delta i plus phi.

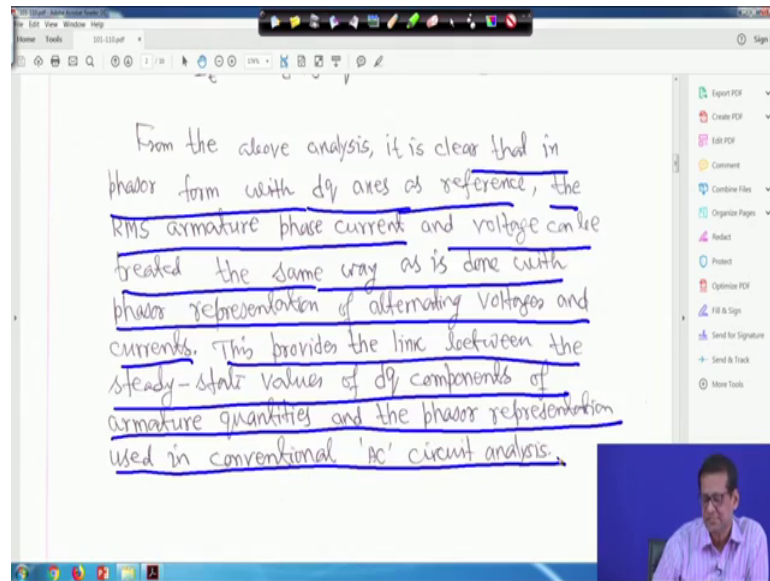
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Therefore, I here I am writing therefore, i_d is equal to your I_t then this one will be $\cos 90$ degree minus delta i plus phi, right. So, that is nothing but your I_t it will be $\sin \delta_i$ plus phi, right that is i_d . Similarly, i_q will be $I_t \cos \delta_i$ plus phi. So, that is what it has been written here, right. So, this is i_d is equal to $I_t \sin \delta_i$ plus phi and i_q is equal to $I_t \cos \delta_i$ plus phi this is equation 161 and this is 162.

Now, and I_t tilde is i_d plus $j i_q$ this is equation 163 because it is complex now, right. So, with this way you can represent I_t is equal to I_t tilde the phasor i_d plus $j i_q$ this way you can represent.

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From the above analysis, it is clear that in phasor form with dq axes as reference, the RMS armature phase current and voltage can be treated the same way as is done with phasor representation of alternating voltages and currents. This provides the link between the steady-state values of dq components of armature quantities and the phasor representation used in conventional 'ac' circuit analysis.

From the above analysis it is clear that in phasor form, right that is in the your what you call in the phasor form with dq axis as reference the RMS armature phase current and voltage can be treated the same way as is done with phasor representation of alternating voltage and current. So, this way we can make it the way we represent the phasor quantities in you know in that your what you call in AC. This provides the link between the steady state values of dq components of armature quantities and the phasor representation used in conventional AC circuit analysis, right. So, this actually dq your conventional AC circuit analysis and in the dq component, this actually there is a link, right, this actually provides the link.

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The relationships between dq components of armature terminal voltage and current are defined by eqns. (137), (138), (140) and (141). Thus

$$e_d = -\omega_r \psi_q - R_a i_d$$

$$\psi_q = L_q i_q$$

$$X_q = \omega_r L_q$$

$$\therefore e_d = X_q i_q - R_a i_d$$

$$e_q = \omega_r \psi_d - R_a i_q$$

So, that means, the relationship between dq component of armature terminal voltage and currents are defined by your this equation 137, 138, 140 and 141. That means, my e_d can be written as minus $\omega_r \psi_q$ minus $R_a i_d$, right.

Similarly, or ψ_q we know from ψ_q is equal to $L_q i_q$. So, we can write ω_r that is your ψ_q is equal to $L_q i_q$ that that has been substituted here. So, $\omega_r L_q i_q$ minus $R_a i_d$ and $\omega_r L_q$ actually is nothing, but X_q . So, X_q actually X_q is equal to $\omega_r L_q$. So, basically e_d which actually $X_q i_q$ minus $R_a i_d$, right, so this is actually equation 164, right.

(Refer Slide Time: 11:27)

$$\therefore e_d = X_q i_q - R_a i_d \quad \dots (164)$$

$$e_q = \omega_r \psi_d - R_a i_q$$

$$\therefore e_q = -X_d i_d + X_{ad} i_{fd} - R_a i_q \quad \dots (165)$$

The reactances X_d and X_q are called the direct and quadrature-axis synchronous reactances respectively.

They represent the inductive effects of the armature mmf wave separately accounting

Similarly, your e_q is equal to $\omega_r \psi_d$, right. So, $\omega_r \psi_d$ minus $R_a i_q$. Now, from the ψ_d expression if you substitute here you will get e_q is equal to your minus $X_d i_d$ then plus $X_{ad} i_{fd}$ minus $R_a i_q$. You put that previously we have derived the ψ_d expression just you put it here and you will get this one. And this is my equation this is our equation 165, right.

Now, the reactances X_d and X_q are called the direct and quadrature axis synchronous reactances respectively. So, X_d is the direct axis and X_q is the quadrature axis synchronous reactances, right. Little bit exercise you do, just put this i_d expression previously derived and you will get this one.

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They represent the inductive effects of the armature mmf wave by separately accounting for its d- and q-axis components.

We have not yet developed a means of identifying the d- and q-axis positions relative to \tilde{E}_t . In order to assist us in this regard, let us define a voltage \tilde{E}_q as:

Now, they represent the inductive effects, right. They represent the inductive effects of the armature mmf wave by separately your accounting for its d and q-axis components, right.

Now, this is what is X_d and X_q . Now, we have not yet developed a means of identifying the d and q-axis positions relative to \tilde{E}_t , right. In order to assist us in this regard let us define a voltage \tilde{E}_q that means, we were defining a voltage say \tilde{E}_q , right.

(Refer Slide Time: 13:03)

(104)

$$\tilde{E}_q = \tilde{E}_t + (R_a + jX_q) \tilde{I}_t$$

$$\therefore \tilde{E}_q = (E_d + jE_q) + (R_a + jX_q) (-I_d + jI_q)$$

Substitution of eqn (164) and (165), followed by reduction of the resulting expression, yields the following expression for \tilde{E}_q in phasor form with d, q axes as reference:

That means we are defining say e_q tilde is equal to E_t tilde plus R_a plus $j \times X_q$ into I_t tilde this you are defining say. Now, E_t is equal to E_t tilde is equal e_d plus $j e_q$ therefore, e_q tilde is equal to e_d plus $j e_q$ plus R_a plus $j \times X_q$ into i_d plus $j i_q$ this is equation 166, right.

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The screenshot shows a presentation slide with the following content:

$$\tilde{E}_q = \tilde{E}_t + (R_a + jX_q) \tilde{I}_t$$

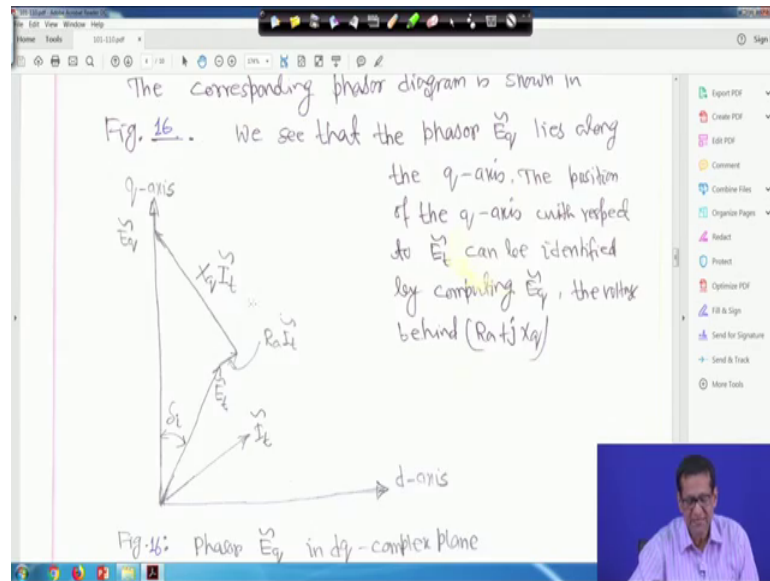
$$\therefore \tilde{E}_q = (e_d + j e_q) + (R_a + jX_q)(i_d + j i_q) \quad \dots (166)$$

Substitution of eqns (164) and (165), followed by reduction of the resulting expression, yields the following expression for \tilde{E}_q in phasor form with d, q axes as reference:

$$\tilde{E}_q = j [X_{ad} i_{fd} - (X_d - X_q) i_d] \quad \dots (167)$$

So, now substituting equation 164 and 165 followed by reduction of the your just hold on followed by the your reduction of the resulting expression it gives actually following expression for E_q tilde in phasor form with dq axis as reference, right. Therefore, in equation 164 and 165 you substitute, right equation of 164 and 165 that means this one, that means, this one that is your equation 164 e_d expression and e_q expression and then you simplify you substitute and you simplify.

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If you simplify, right then you will get \tilde{E}_q will be j into $X a + I d$ minus in bracket $X d$ minus $X q$ into $I d$ this is equation 167. That means, it is a pure complex quantities coming that this \tilde{E}_q will lie on the q-axis, right. Therefore the corresponding phasor diagram is shown in figure 16. So, this is \tilde{E}_q because it will lie on the q-axis because only j no real part is involved, right. So, it is j . So, this is \tilde{E}_q is equal to \tilde{E}_t plus $R a + j X q$, right. So, whatever we have assume here \tilde{E}_q is equal to \tilde{E}_t plus $R a + j X q$.

So, this is the phasor diagram. And I_t is lagging from \tilde{E}_t by an angle ϕ not shown here this is \tilde{E}_t and this is I_t this is d-axis and this is q-axis, right. Therefore, the corresponding phasor diagram is shown in figure 16, this is figure 16, right. So, we see that the phasor \tilde{E}_q lies along the q-axis. The position of the q-axis with respect to \tilde{E}_t can be identified by computing \tilde{E}_q the voltage behind that is $R a + j X q$, right so that means, your this one $R a + j X q$, right.

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Rotor Angle

Under no-load or open circuit conditions,
 $i_d = i_q = 0$. Substituting in eqns. (137), (138),
 (140) and (141) yields,

$$\psi_d = L_{ad} i_{fd} \quad \checkmark$$

$$\psi_q = 0 \quad \checkmark$$

$$e_d = 0 \quad \checkmark$$

$$e_q = X_{ad} i_{fd} \quad \checkmark$$

Therefore

Next is rotor angle under no load or open circuit conditions i_d is equal to i_q is equal to 0, right therefore, substituting equation 137, 138, 140 and 141. This will give because under no load or open circuit condition i_d and i_q both are 0 you substitute that in those equation in your in equation your 137, 138, 140 and 141 you substitute that. If you do so, ψ_d will be is equal to L_{ad} into i_{fd} , ψ_q will be 0, e_d will be 0, and e_q will be $X_{ad} i_{fd}$, right, this will get.

(Refer Slide Time: 16:23)

Therefore,

$$\vec{E}_t = e_d + j e_q$$

$$\therefore \vec{E}_t = j X_{ad} i_{fd} \quad \dots (168)$$

Under no-load conditions, \vec{E}_t has only the q-axis component and hence $\delta_i = 0$. As the machine is loaded δ_i increases. Therefore, the angle δ_i is referred to as the internal rotor angle or load angle. The relationship between power output and the rotor angle is nonlinear and is of fundamental importance in power system stability studies.

Therefore \vec{E}_t we know $e_d + j e_q$, but e_d is equal to 0 therefore, \vec{E}_t will

become jX_{ad} into i_{fd} , this is equation 168, right. So, under no load conditions e_t has only the q-axis component and hence δ is 0, right. So, just hold on, right. As the machine is loaded δ increases. Therefore, the angle δ is referred to as the internal rotor angle or load angle of the machine that you studied in synchronous machine, right. So, as the machine is loaded δ increases therefore, as the angle δ refer to as the internal rotor angle or load angle of the machine. The relationship between power output and the rotor angle is non-linear and is of fundamental importance in power system stability studies, right.

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(16)

The angle δ_i represents the angle by which the q-axis leads the stator terminal voltage phasor \vec{E}_t , and it is given by

$$\delta_i = 90^\circ - (\alpha - \theta_0) \quad \dots (169)$$

where α is the phase angle of E_a and θ_0 is the value of θ with respect to the time axis.

Therefore δ_i is the angle between the

The angle δ_i actually represent the angle by which the q-axis lead the stator terminal voltage phasor E_t and it is given by that I told you earlier the δ_i will be 90 degree minus α minus your θ_0 , right. So, this is actually your what you call that your δ_i . So, if you go back to that previous figure 16 you see the δ is equal to 90 degree minus α minus θ_0 . This is equation 169, right.

(Refer Slide Time: 18:01)

phasor E_t , and it is given by

$$\delta_i = 90^\circ - (\alpha - \theta_0) \quad \dots (16g)$$

where α is the phase angle of e_a and θ_0 is the value of θ with respect to the time origin.

Therefore, δ_i depends on the angle between the stator and rotor magnetic fields.

For any given machine power output, either α or θ_0 may be arbitrarily chosen, but not both.

Now, where alpha, where alpha is the phase angle of e a and theta 0 I told you before is the value of theta with respect to the time origin set is equal to 0, right. Therefore, delta i depends on the angle between the stator and rotor magnetic fields, right.

For any given machine power output either alpha or theta 0 may be arbitrarily chosen, but not both your project either alpha beta to 0 for any main forum for any given power output, but not both you have to choose either alpha or theta 0 for any for for any given power output, but not both, right. So, that is the condition.

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Steady-State Equivalent Circuit

If saliency is neglected,

$$X_d = X_q = X_s$$

Where X_s is the synchronous reactance.

Therefore,

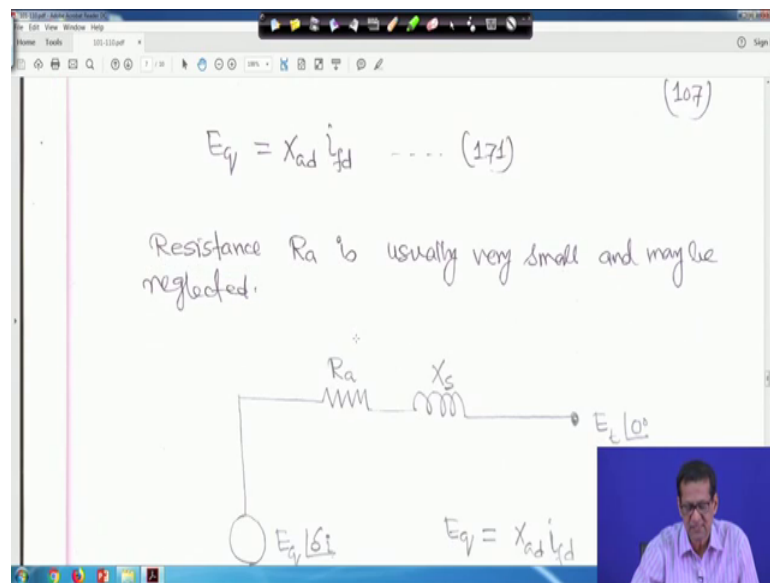
$$E_q = E_t + (R_a + jX_s) I_t \quad \dots (170)$$

With $X_d = X_q$, from eqn. (16g), the mag of E_q is given by

Now, steady state equation a steady state equivalent circuit. If saliency is neglected then we can assume X_d is equal to X_q is equal to X_s , right if you neglect the saliency then X_d is equal to X_q is equal to X_s , right, where X_s is the synchronous reactants we are assuming both are same.

Therefore we can write E_q we know E_t plus R_a plus j your it your X_q , so now, X_q is equal to X_s . So, $j X_s$ into I_t this is equation 170, right where X_t is equal to X_q . So, from equation 167 the magnitude of E_q is given by.

(Refer Slide Time: 19:21)



So, this is then E_q will be $X_{ad} i_{fd}$. So, that is from equation 167 I mean here.

(Refer Slide Time: 19:33)

(140) and (141) yields,

$$\Psi_d = L_{ad} i_{fd}$$

$$\Psi_q = 0$$

$$e_d = 0$$

$$e_q = X_{ad} i_{fd}$$

Therefore,

$$\vec{E}_t = e_d + j e_q$$

This is your E_q is equal to $X_{ad} i_{fd}$, right this is this is your just hold on 167, right because X_d is equal to X_q , here X_d is equal to X_q . So, is equal to X_s , so X_d it is this term will be not, will not be there. So, it will be only j into $X_{ad} i_{fd}$, but if you take the magnitude then E_q will be only is equal to $X_{ad} i_{fd}$ only magnitude. So, that means, your this is my capital E_q is equal to $X_{ad} i_{fd}$ this is equation 171, right.

(Refer Slide Time: 20:15)

Resistance R_a is usually very small and may be neglected.

ω

$E_q \angle \delta$

i_d

$E_t \angle \theta$

R_a

jX_s

$E_t \angle \theta$

I_t

$E_q \angle \delta$

$E_q = X_{ad} i_{fd}$

$X_d = X_q = X_s$

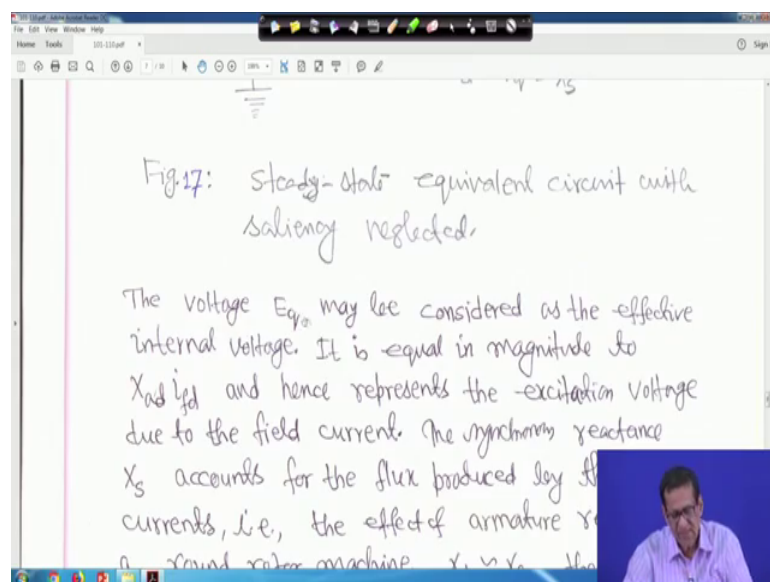
Fig.17: Steady state equivalent circuit

Therefore resistance R_a is usually very small and maybe neglected therefore, this equivalent circuit we can make that this is my E_q angle δ this is $R_a X_s$ and this is

$E_t \angle 0$. Only one thing I have to make it that this is the current then this is a I_t tilde, right. Therefore, E_q is equal to X_{ad} i_f magnitude one and X_d is equal to X_q and X_s this is $E_t \angle 0$ and this is we have taken $E_q \angle \delta_i$ that means, I mean if you take this one as reference. So, E_q leading E_t by an angle δ_i because this was this was your q-axis and this is your E_q , and this was your E_t , and this angle was your δ_i , right. So, basically E_q leading E_t by an angle δ_i , if E_t has a reference. So, $E_t \angle 0$, right that is why we have written $E_t \angle 0$ and $E_q \angle \delta_i$ and R a axis, right.

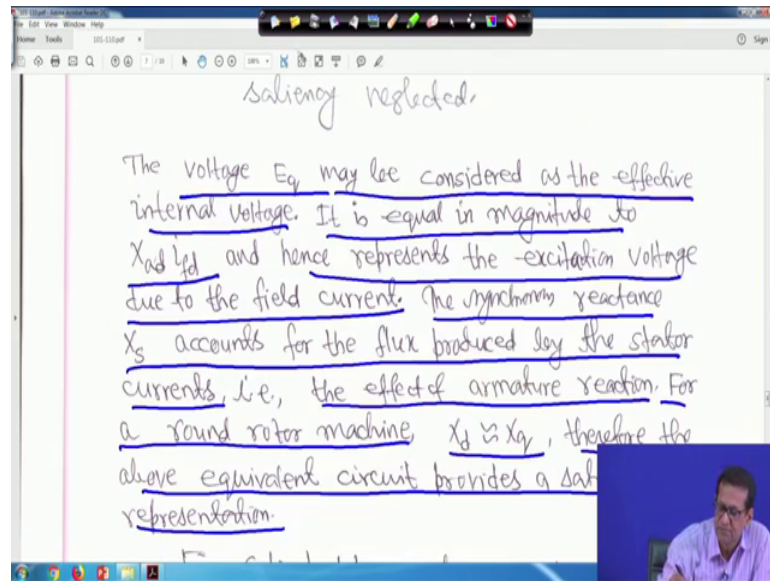
If you want to put j you can put j here, right. But generally for machine R a is very small can be neglected, but this is the equivalent steady state equivalent circuit, right.

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So, the voltage E_q may be considered as the effective internal voltage, right.

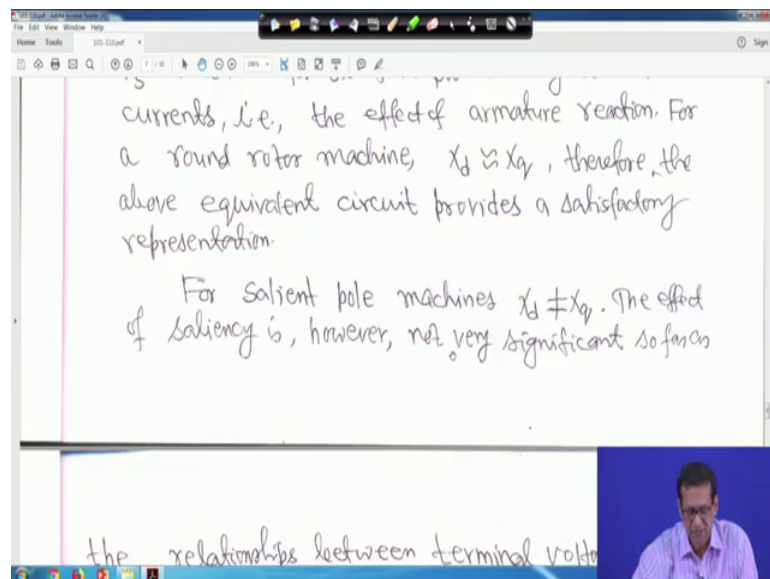
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So, the voltage E_q may be considered as the effective internal voltage, right. It is equal in magnitude to $X_d i_f d$ that we have seen just now. And hence represent the excitation voltage due to the field current because it is $X_d i_f d$ $i_f d$ is the field current, right.

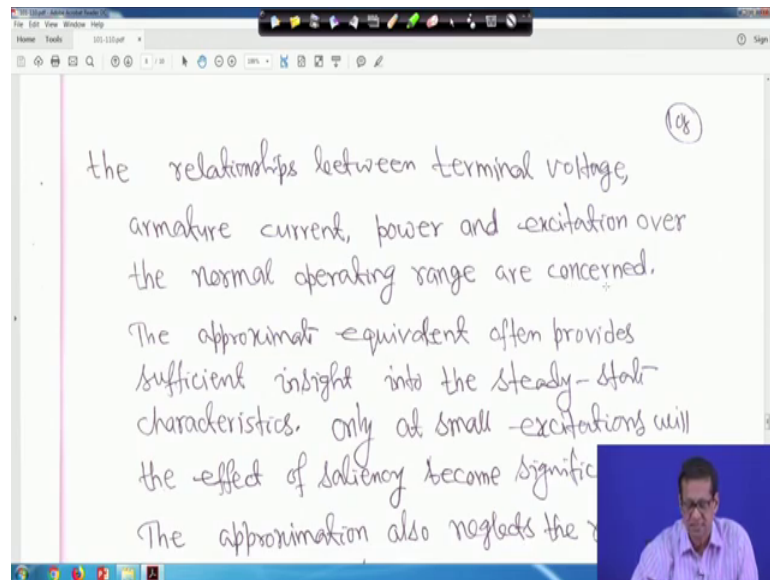
The synchronous reactance X_s accounts for the flux produced by the stator currents that is the effect of armature reaction, right for a round rotor machine, right X_d approximately is equal to X_q is equal to X_s we have taken. Therefore, the above equivalent circuit provides a satisfactory representation, right.

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For salient pole machine of course, X_d is not equal to X_q , the effect of saliency is however, not very significant, right, so far as the relationship between terminal voltage, armature current, power and excitation over the normal operating range are concerned, right.

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The approximate equivalent you call often provides sufficient insight into the steady state characteristics. Only at small excitations will the, that is will the effect of saliency becomes significant. The approximation also regards the your what you call neglects the your reluctance torques due to saliency, right.

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the relationships between terminal voltage, armature current, power and excitation over the normal operating range are concerned. The approximate equivalent often provides sufficient insight into the steady-state characteristics. Only at small excitations will the effect of saliency become significant. The approximation also neglects the torque due to saliency.

Now, next is active and reactive power.

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Active and Reactive power.

$$S = \tilde{E}_t \tilde{I}_t^* = (e_d + j e_q)(i_d - j i_q)$$

$$\therefore P_e + j Q_e = (e_d i_d + e_q i_q) + j(e_q i_d - e_d i_q)$$

$$\therefore P_e = e_d i_d + e_q i_q \quad \dots (172) \quad \checkmark$$

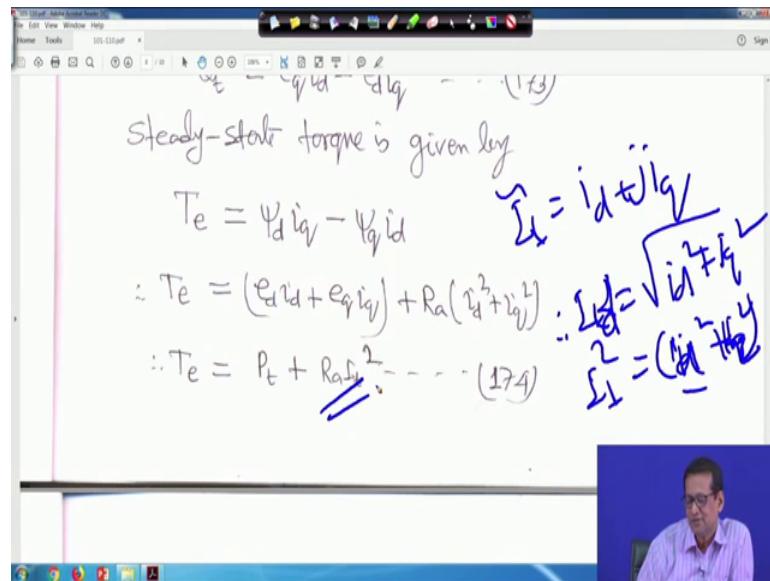
$$Q_e = e_q i_d - e_d i_q \quad \dots (173) \quad \checkmark$$

Now, we know S is equal to we know in general from load flow studies we know that either when you, right using general, right that $P - jQ$ is equal to b conjugate i or $P + jQ$ is equal to $b i$ conjugate, right. So similarly active and reactive power, right, S the apparent power $\tilde{E}_t \tilde{I}_t^*$. So, \tilde{E}_t is equal to $e_d + j e_q$ and \tilde{I}_t is conjugate \tilde{I}_t is equal to $i_d - j i_q$. So, \tilde{I}_t^* is $e_d + j i_q$ that is $e_d - j i_q$, right sorry $i_d - j i_q$. So, $e_d + j i_q$ into i_d

minus $j i q$ therefore, at S is equal to your, S is actually p_t plus $j Q_t$ because it is $e i$ conjugate, right. Therefore, E_t you can if you multiply it will be $e d i d$ plus $e q i q$ plus $j e q i d$ minus $e d i q$.

Now, we separate real and imaginary part. So, p_t will be $e d i d$ plus $e q i q$ and Q_t will be $e q i d$ minus $e d i q$. This is equation 172, this is equation 173, right.

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Now, steady state torque is given by T is equal to this we have derived that $\psi_d i q$ minus $\psi_q i d$ this we have derived or simply this one if you just put those ψ_d ψ_q all these expression you will get T_e is equal to this one little bit I ask you to derive of your own, right.

So, just put those expression you will get T is equal to $e d i d$ plus $e q i q$ plus R_a into $i d$ square plus $i q$ square or T is equal to this is my power. Just now we have seen $e d i d$ plus $e q i q$ this is my this is my power p_t , right just now we have seen and this is $R_a I_t$ square because your I_t is equal to e sorry just hold on I_t is equal to $i d$ plus $j i q$. Therefore, your magnitude if I take magnitude of this one is equal to this one, right therefore, my I_t square is equal to your $i d$ square plus $i q$ square, right that is what is written here $R_a I_t$ square. This is your equation 174.

So, now question is that your what you call little bit more steady state computation will go in the next class, next lecture. So, now, if you look into that that T is equal to equal to

P_t plus $R_a I^2$.

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$$P_t = \psi_d i_d + \psi_q i_q \quad \dots (172)$$

$$Q_t = \psi_q i_d - \psi_d i_q \quad \dots (173)$$
 Steady-state torque is given by

$$T_e = \psi_d i_q - \psi_q i_d \quad \underline{T_e \approx P_t}$$

$$\therefore T_e = (\psi_d i_d + \psi_q i_q) + R_a (i_d^2 + i_q^2)$$

$$\therefore T_e = P_t + R_a I^2 \quad \dots (174)$$

Now, in general if you look into this that this, this part, this part $R_a I^2$ actually very small, right. Now, if you do so, that means, if you do so that this is actually torque will be approximately equal to the power output P_t , right. Therefore, in per unit system when you convert into per unit system, if I generally if you neglect the loss you will find in per unit torque and power both are same because this term $R_a I^2$ is very small, right. So, many-many in our analysis many cases we assume it is a you assume that per unit torque is equal to per unit power, if you convert it to this thing because this term is very small. Therefore, torque and whatever we have made it this torque is equal to this one P_t plus $R_a I^2$ everything is in per unit and this term is very small compared to this term therefore, torque is approximately in power in per unit, right.

So, only thing is that that after this I mean after this we have one more thing that is your just computational procedure, and we will take one example little bit more derivation is there. And after that we will go for the your what you call that your dynamics that swing equation step by step we will follow, right and slowly and slowly we will go you know you know into much deeper analysis particularly in that your Laplace domain and we will derive all these all these your mathematical model block diagram representation.

And later we will see that stabiliser or system stabiliser your (Refer Time: 27:48) stabiliser we will consider and slowly an Eigen value analysis and participation factor all

these things will your what you call will examine, but that is in you know slowly and slowly we move into that. But after this we will go little bit of computational analysis and then we will come to the swing equation.

Thank you very much. We will back again.