

Network Analysis
Prof. Tapas Kumar Bhattacharya
Department of Electrical Engineering
Indian Institute of Technology - Kharagpur

Lecture – 22
R-L Circuit with Sinusoidal Excitation

Welcome to lecture number 22nd and we have been solving on circuits, where this supply voltage will be time varying and as well as there will be energy storing elements in this circuit and we have considered RL circuit, RC circuit, first with DC sources and then the basic steps will be write down if you are interested to find out the current flowing through an inductor, first develop a differential equation involving that current.

And note down the initial conditions that is i_0 minus which we will give you i_0 plus or i_0 which will be same in general therefore, then you write down the differential equation but after that do not integrate to get the solution. See, in our last few lectures I wanted to stress upon this point; if the differential equation is known, you should straight away write down the solution of the differential equation which will have a natural response part.

That is if the right hand side was 0, no excitation, what would be the solution; that solution depends upon the characteristic roots and exponential terms will appeared there, of course with some constants and there will be another solution corresponding to forcing function or excitation present. One important thing to be noted is this; this forcing function solution alone will satisfy the natural response.

So, we consider several problem today, I will consider another problem and show you how to apply the same technique, when the excitation is say a sinusoidal voltage, okay.

(Refer Slide Time: 02:28)

lec-22

R-L series ckt excited from a sinusoidal voltage source

$i(0) = 0$ let $v_s(t) = V_m \sin \omega t$

$L \frac{di}{dt} + Ri = v_s(t) = V_m \sin \omega t$

or $\frac{di}{dt} + \frac{R}{L}i = \frac{V_m}{L} \sin \omega t$

$(D + \frac{R}{L})i = 0 \rightarrow \text{Ch. } e^{mt}$
 $m + \frac{R}{L} = 0$
 $m = -\frac{R}{L}$

will satisfy alone

$i(t) = A e^{-\frac{R}{L}t} + (K_1 \sin \omega t + K_2 \cos \omega t)$

$i(t) = A e^{-\frac{R}{L}t} + K_1 \sin \omega t + K_2 \cos \omega t$

natural solⁿ solⁿ due to forcing fⁿ

Indentity

$K_1 \omega \cos \omega t - K_2 \omega \sin \omega t + \frac{R}{L} K_1 \sin \omega t + \frac{R}{L} K_2 \cos \omega t = \frac{V_m}{L} \sin \omega t$

$\frac{R}{L} K_1 - \omega K_2 = \frac{V_m}{L}$
 $\omega K_1 + \frac{R}{L} K_2 = 0$

For example, consider a RL circuit; RL series circuit excited from a sinusoidal voltage source and so the thing is like this R here inductance in series L and here is a source which is time varying, v_t is the voltage source and $i(0)$ minus this is the current I want to solve; i as the function of time and $i(0)$ minus is equal to 0, it is suppose given which was open for a long time and so on, now this switch s is closed at t equal to 0.

Then, how the current will grow in the circuit or will flow in the circuit as a function of time, this I want to find out, then the first step is write down the voltage drops that is Ri here with proper polarity and this is $L \frac{di}{dt}$ and we know these that the sum of this 2 voltages; $L \frac{di}{dt}$ plus Ri is equal to your input v_t . Now, let v_t , the applied voltage is V_{max} into some $\sin \omega t$, suppose, so applied voltage is $\sin \omega t$.

That means I have close this switch, suppose if I sketch the supply voltage here, this the voltage I am applying, this is your V_{max} and this axis is ωt , okay and this is 0, π , etc., so I am closing the switch, this axis is ωt , means also t , ω is constant supply frequency, so this axis also determines the time axis in some other scale divided by ω . So, at t equal to 0, I am closing the switch, so I am closing the switch at this point; s is closed at 0.

And that is why I am describing this as $V_{max} \sin \omega t$ and I want to get the solution, so I; let me be systematic rules I know, what should I do, so this is equal to $V_m \sin \omega t$, is it not.

Then, I will write it fast as divide by L; $\frac{di}{dt} + \frac{R}{L}i$ is equal to $\frac{V_m}{L} \sin \omega t$, this is the differential equation. What is the characteristic equation? Characteristic equation is $D + \frac{R}{L}i$ D is $\frac{d}{dt}$ is equal to 0 if you do, you get the characteristic equation, is it not.

And D is being an operator, so characteristic equation is $m + \frac{R}{L}$, it is a first order system, only one characteristic equation, only one energy storing elements, so m is equal to minus $\frac{R}{L}$, therefore the solution now comes the interesting part; the solution I will write straight away, it will be equal to some constant A into e to the power mt that is minus $\frac{R}{L}$ into t that will be the first part.

And the second part will be as we know the; it should be since it is Vt , it should be like $k_1 V$ plus k_1 is a bar, k_2 , another constant $V \dot{}$ plus k_3 , another constant $V \ddot{}$ and up to infinity, this is a solution due to forcing function, so solution due to forcing function is a linear combination of the excitation and its higher order derivative. Now, here this you see, it is $\sin \omega t$, so it should be if you do this, so you have to go on differentiating this $\sin \omega t$, so this will give rise to which I am not going to write.

What I am telling; it will be some constant into $\sin \omega t$, if you put this plus if you differentiate this, it will be $\omega \cos \omega t$, is it not, that ω will multiply with k_2 bar to be another constant, then if you differentiate $V \dot{}$ to get $V \ddot{}$, so once again $\sin \omega t$ term will come, so alternatively the terms will be $\sin \omega t$, $\cos \omega t$, of course with different constants here and that is to be added up, up to infinity.

Now, therefore it looks like; there will be terms like, so this is equal to A into e to the power minus $\frac{R}{L}$ into t plus it looks like, there will be $\sin \omega t$; a number of infinite number of $\sin \omega t$ terms, so whatever it is I will call that thing to be some big constant into $\sin \omega t$, is it not, this I can always do because those coefficients are constant. Similarly, there will be infinite number cosine ωt terms.

And that I will group those; if you take cosine ωt out, this will be another constant term, of course C and D will have that ω terms but ω is constant, not circuit parameter decided by supply frequency that is all but this will be the solution, so you see I get the solution of this network, the form of the solution at least I do not require any time, okay, find out the characteristic root, this is the natural part, this is this part, okay.

Now, if you wish I was writing it as some big K_1 and K_2 , okay, generally, for forcing function the constants I was writing as K_1 , K_2 , you can do that. Now, to find out K_1 and K_2 , first I will find out K_1 , K_2 , so this is the this part, solution due to forcing function and this part is natural solution, got the point. Now, to so this forcing function alone satisfies this equation, this solution, this part will satisfy this; will satisfy alone.

Because total solution satisfies this but natural response if you substitute this will give rise to 0, so I can very well say except t alone, will also satisfy that, that is the key of this whole thing, so I will put this here, so it is $D_i dt$, so forcing function alone will satisfy this, so it will be $K_1 \omega \cos \omega t$, I am differentiating this $D_i dt$, so $K_1 \omega \cos \omega t$, then minus I have to differentiate this also, so $K_2 \omega \sin \omega t$, $\cos \omega t$ if you differentiate, it gives you ωt term, is it not, that is okay.

So, this gives you $D_i dt$ plus R by L into this current due to forcing function which will be equal to $K_1 \sin \omega t$ plus R by L ar because R by L is common into i ; R by L into $K_2 \cos \omega t$ and I am telling this will be equal to B_m by $L \sin \omega t$, is it not, this is the thing. So, I am determining first the constants K_1 , K_2 ; K_1 , K_2 can be determined and as I told you, so there are 2 unknown not infinite numbers of unknowns.

And these 2 unknowns can be solved for by having 2 equations, so and this is an identity, mind you, this is an identity, therefore coefficient of $\sin \omega t$ on both the sides will be equal, so what is the coefficient of $\sin \omega t$; both the sides it will be $\sin \omega t$ terms are here, this is the $\sin \omega t$ terms, this is the $\sin \omega t$ terms, therefore since it is an identity I will write R by L into K_1 minus ω into K_2 .

And on the right hand side, there is a sin omega t, whose coefficient is Vm by L, must be equal to Vm by L, that is what I am telling, this is one equation and the second equation is coefficient of cosine omega t, equate, so it will be omega K1; this one and another cosine omega t term is there plus R by L into K2, I hope signs are all okay know, this is, this must be equal to; there is no cosine term here must be equal to 0.

These are the 2 equations I have to solve, so these 2 equations I have to solve, only thing it is Vm that I will write.

(Refer Slide Time: 15:58)

Handwritten derivation showing the solution for the current $i(t)$ in an RL circuit. The equations are:

$$\begin{aligned} \frac{R}{L} k_1 - \omega k_2 &= \frac{V_m}{L} \\ \omega k_1 + \frac{R}{L} k_2 &= 0 \end{aligned}$$

Solving for k_1 and k_2 :

$$k_2 = -\frac{L\omega}{R} k_1$$

$$\frac{R}{L} k_1 + \frac{L\omega^2}{R} k_1 = \frac{V_m}{L}$$

$$\therefore k_1 = \frac{V_m/L}{\frac{R}{L} + \frac{\omega^2 L}{R}} = \frac{V_m}{R + \omega^2 L} = \frac{R V_m}{R^2 + \omega^2 L^2}$$

Then $k_2 = -\frac{\omega L}{R} \cdot \frac{R V_m}{R^2 + \omega^2 L^2} = -\frac{\omega L V_m}{R^2 + \omega^2 L^2}$

The current is given by:

$$i(t) = A e^{-\frac{R}{L} t} + \frac{R V_m}{R^2 + \omega^2 L^2} \sin \omega t - \frac{\omega L V_m}{R^2 + \omega^2 L^2} \cos \omega t$$

Using the identity $\sin \alpha \cos \beta = \frac{1}{2} [\sin(\alpha + \beta) + \sin(\alpha - \beta)]$, the expression can be written as:

$$i(t) = A e^{-\frac{R}{L} t} + \frac{V_m}{\sqrt{R^2 + \omega^2 L^2}} \left[\frac{R}{\sqrt{R^2 + \omega^2 L^2}} \sin \omega t - \frac{\omega L}{\sqrt{R^2 + \omega^2 L^2}} \cos \omega t \right]$$

The phase angle θ is determined by:

$$\tan \theta = \frac{\omega L}{R}$$

The final expression for the current is:

$$i(t) = A e^{-\frac{R}{L} t} + \frac{V_m}{\sqrt{R^2 + \omega^2 L^2}} \sin(\omega t - \theta)$$

A is to be determined from boundary condition $i(0) = 0 = i(\theta)$

So, previous 2 equations are this one, so I have to solve for them, so from this equation I will say that K2 is equal to from this equation, R by L K2 is equal to minus omega K1 and it will be L by R, this is K2 and this thing I will submit it, put it there, so I am going to find out K1, so it will be R by L into K1 was there from this equation minus omega K2, so it will become plus L omega square by R into K1 and that I am telling is equal to Vm by L.

So, this is the thing therefore, K1 will be equal to Vm by L in the numerator and below it will be R by L, plus omega square L by R, is it not, this thing I have brought it down which will be equal to Vm by this L you bring down, it will be equal to R, L cancels out plus omega square L square divided by R and these are also can be put on the top, so R into Vm, it will be equal to R square plus omega square L square.

So, this is the value of K_1 and if this is K_1 , then K_2 ; K_2 is minus ωL by R into K_1 and K_1 is this one; $R V_m$ by R^2 plus $\omega^2 L^2$, so R goes and you will be left with minus ωL divided by R^2 plus $\omega^2 L^2$ into V_m , this will be the thing, therefore the solution; total solution will be it will be equal to $A e^{-R/L t}$ that is the natural response plus solution due to forcing function.

So, K_1 into $\sin \omega t$ and k_2 is this one into $\sin \omega t$ plus K_2 , which happens to be a negative thing like this, so ωL by R^2 plus $\omega^2 L^2$ into V_m into $\cos \omega t$, so this is this total solution, only thing is this can be manipulated a bit and it can be written in this form that plus this one and here what you do; you take V_m by square root of R^2 plus $\omega^2 L^2$ common from this 2 terms.

Then you will be left with R over this thing, $\sin \omega t$ and minus ωL and V_m I have taken outside, 1 square root I have taken, so another square root also is left here, $\cos \omega t$, it will be like that. Now, these 2 things you see this term square and this term square gives you 1, is it not, this square plus this square, they will give 1, therefore I can define let R by; I am sorry, so R ; let R I defined as R by root over R^2 plus $\omega^2 L^2$ as $\cos \theta$.

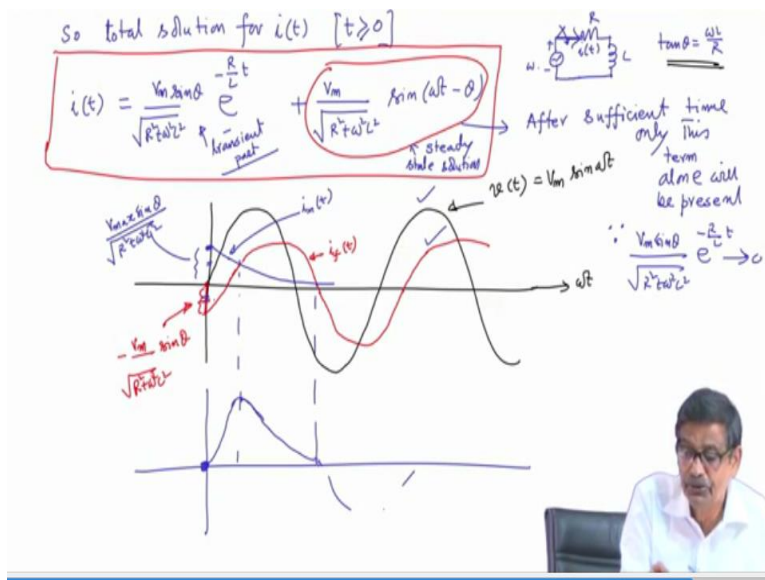
And then ωL by this one, same square root has to be $\sin \theta$ why because this square plus this square is 1, therefore these equation, so this is $\cos \theta$ and this term is $\sin \theta$, so I have $\sin \omega t \cos \theta$ minus $\cos \omega t \sin \theta$, therefore this solution is nothing but $A e^{-R/L t}$ plus V_m by this quantity and this one can be written as $\sin(\omega t - \theta)$, got the point.

So, this is it, this is the solution now, in this solution, still A is to be determined, capital A has not been determined yet, mind you, after you write down the total solution, then you apply the boundary condition for example, A is to be determined from boundary conditions which is given to be i_0 minus equal to 0 that is equal to i_0 , so I put that and find out the constant here; so here there is space I will do here that thing.

So, put t equal to 0 on both sides, so it will be equal to 0 is equal to i0 is 0 is equal to A e to the power 0 is 1 and in this expression t is equal to 0, so it will be plus Vm over root over R square plus omega square L square sin of minus theta. What is theta; theta is tan theta is equal to as you can see I have defined cos theta, sin theta, so tan theta is omega L by R. so, sin theta known in fact.

Therefore, A will be equal to Vm divided by root over R square plus omega square L square sin minus theta is minus sin theta, so this becomes plus then, if you change side, got the point, so this is the value of A.

(Refer Slide Time: 25:11)



Therefore, total solution I will write in the next page as so, total solution for it, I mean I want to get the solution t greater than equal to 0 will be equal to it is equal to some e to the power minus R by L into t plus solution due to forcing function V max over this thing into sin omega t minus theta and here the value of the constant was found out to be this thing; V max by this into sin theta, so this constant is V max by root over R square plus omega square L square sin theta.

Mind you, theta is constant, once this circuit is known, R, L, and your supply frequencies omega, then you know and we are solving for this current and we know that tan theta is equal to omega L by R, so this will be the complete solution. Let us see whether this solution is okay or not; at t

equal to 0, i_0 must be 0. So, at t equal to 0, this is B_m by this into \sin minus θ and it is $V_m \sin \theta$ by this.

So, \sin minus θ is minus $\sin \theta$, so this 2 term will cancel out and will you the result, so this is the total solution, got the point, so if you sketch as I told you, it is always better you sketch the solution, so sketching for the solution, you should do it like this; first sketch v_t , this one is v_t which is equal to $V_m \sin \omega t$, here is ωt and then the current waveform is this kind of thing.

See, if you look at this current waveform, it has got 2 distinct part; one part is V_m by root over R square ω square L square $\sin \omega t$ minus θ , so θ value is known from this, so current waveform will be lagging the voltage waveform by an angle θ , I am drawing it with red, this is the solution due to forcing function, this part is this part, got the point, only this part I have drawn.

Then, there is a part but this is the only ift but there is a natural part, you have to add these waveform to these to get the total solution. Now, what will be that part; that part is telling that this one; the magnitude at t equal 0 of the forcing function is minus V max by root over R square plus ω square L square $\sin \theta$, \sin minus θ , minus you bring outside, so at t equal to 0, this is the current; negative current.

But at t equal to 0, this must give this much of positive current and in fact, it is given, so this I will use a different colour, yeah, so it will this length is equal to this length, these 2 lengths are same and this one is at t equal to 0, V max $\sin \theta$ by root over R square plus ω square L squares and then it is exponentially decaying, is it not, e to the power minus R by L and this sum of this curve plus this curve will give you the total current.

Therefore, initially this sine wave after sometime of course, it will be only this part present, after sufficient time; sufficient time means talk in terms of time constant of the circuit after sufficient time only this term is present, this term alone will be present since V max $\sin \theta$, the other

term by root over $R^2 + \omega^2 L^2$ into $e^{-\frac{R}{L}t}$ will tend to 0 that is what it is.

After 1, 2 or may be 3, 4 cycles it will be only black and red curve will be present, of course initially this current waveform will start has to; if you are try to draw the resultant current below, so this is I will write, it is fair to write, so this if i_1 and this I will write, this blue one is natural response; int. So, the total current will be some of this and it has to start from 0 and then this plus this, this plus this, this plus this, so here it will grow like this, some asymmetry will be there initially.

Then that thing, once again will fall, it will be like this, here it will be 0, I mean something like that, you get then this is vanishingly small, I mean something like that so, initially there will be asymmetry, it will be shifted, it is called decision and then of course, after long time what will remain is this voltage waveform black and this red waveform, int will vanish. So, the; this part is; this is the total solution and we find the in the total solution, there will be a transient part which will decay.

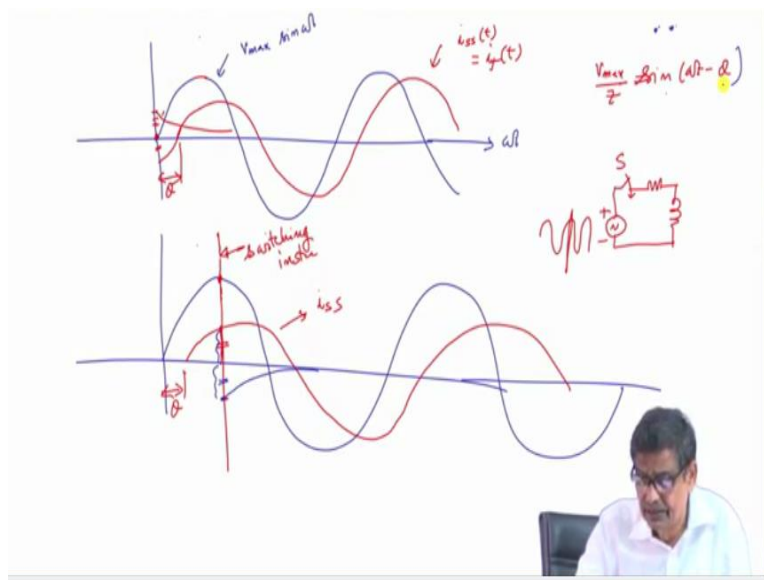
And this part is called steady state solution, this is also called solution due to forcing function is also called steady state solution and this is called the transient part, got the point. So, what is the conclusion? See, behind all this maths, the conclusion is to be understood very carefully, the conclusion is any RL circuit, you suddenly switch on with the help of a switch then, the response of the current will comprise of 2 parts.

But some of this 2 at t equal to 0 has to be 0 that is there, it has to start from 0 and there is an exponential term which will appear and that will go down to 0 and finally, it is the steady state solution which will sustain, any switching of RL circuit with some sinusoidal excitation a transient will be born but that transient will not last long, it will die down, die down by what time; well, it depends upon the time constant of the circuit may be after few cycles, that transient will be all together vanished.

That is why while analysing AC circuit, say RL, RC, this type of circuits, you energised with sinusoidal, people often neglect this transient part, I am interested because with see, one supply cycle. 50 hertz is the supply frequency, 20 milli second is 1 cycle, RL values are such that the order of time constant will be of the order of some few milli seconds maybe 3 cycles that is 60 milli second may be if it is very inducting.

But after that 60 milli second, only this fellow and this fellow will be present, who bothers that transient was there when you switched on the circuit of course, these initial current which is produced which is some of transient and the steady state current, natural and force response, these are very important piece of studies in power system, when you energise a power circuit line have been lot of inductance, then initial current search may be more.

(Refer Slide Time: 37:02)



And last point I want to make it behind all this maths, you should not forget this for example, let me tell you which am not doing but I must point this out, so to create interest, for example you have a supply voltage like this, this point you listen carefully, this is ωt , no max here, $V_{max} \sin \omega t$ is the voltage I am counting time here. Now, the point is that this supply voltage is present like that between to say your supply or to supply, it is supply is present and it is doing like this.

Should I call it $\sin \omega t$ or else depends upon me, if I start counting my time from this instant I will say, okay it may be called $\cos \omega t$ and so on but what I know for certain is that this steady state current will be doing like this, this I can draw, red one is the steady state component; i_{ss} , which is equal to i forcing function t , I just did last time, so circuit; the steady state current is very simple you know that; $V_{max} \frac{1}{\sqrt{R^2 + \omega^2 L^2}}$ is called impedance into $\sin(\omega t - \theta)$ and this is θ , is it not.

Now, what I am telling; I am not going to do any maths but I am telling you one very interesting observation, this if you understand, I will be happy, I mean, you can do several things, for example I say that I have a RL other circuit here, I am going to energise it with this switch and I will close, what is this supply voltage; supply voltage is doing like this, I have not yet called where is t is equal to 0.

But I will tell that when I close this switch, I am not really sure because I cannot see the supply when I switch on say your fan in your home, you switch on, it is after all some sort of RL circuit anyway, I may switch on when this supply voltage was maximum, when the supply voltage was 0, it is left to chances, I do not know really. Now, suppose I say in the previous problem I told you, I have close the switch, when the supply voltage was 0 and going towards positive.

And then I will tell but current through the inductor cannot change instantaneously, so steady state demands this much negative current is flowing, so this much of positive current must appear to make it 0 and that will be the transient part that way I also I can now imagine how things are going. Suppose, I say that with this I will say, this is your supply voltage, I am not writing ωt equal to 0 etc., okay this is your supply voltage.

But steady state current will be displaced from A it by an angle θ ; this angle, is it not, this is the steady state current, red one; i_{ss} , now suppose I say that I will close this switch at this instant, when the supply voltage was around this point okay, here also there will be we expect some transient to come and that will die down but what will be the value of the transient current, right at the (t) (41:28), this is the switching instant, you close this switch, when it was doing like this at that point.

But current through the inductor cannot be 0, steady state demands that the current is this much, positive, do your max, the way I did but I am telling you believe me, your; there will appear a negative current of same amount, here lies the catch, these and these 2 will be equal and this current will decay down to 0, this will be the nature of the transient. So, behind all this mathematical states getting the results, one can easily write down the steady state part of the solution, which is very easy to write; $\sin \omega t - \theta$.

From that decide at whatever time you switch on the circuit, you make up what is, we will continue with this in the next class, thank you.