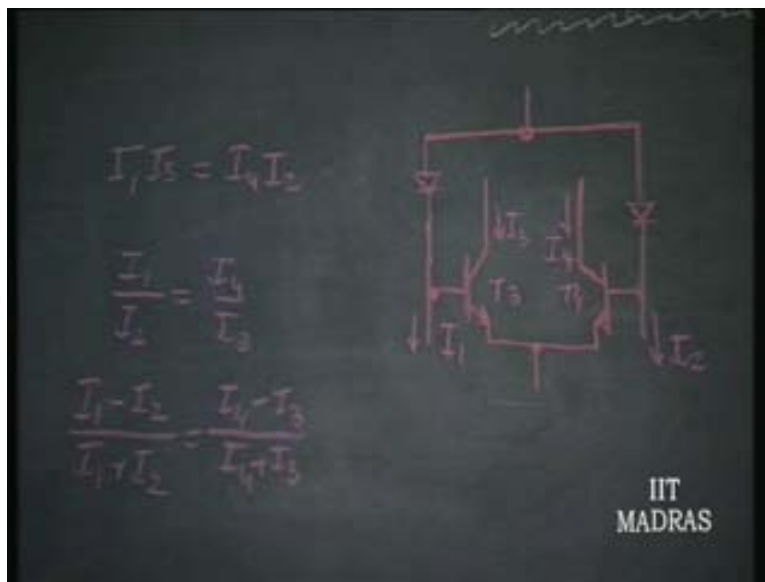


**Analog ICs**  
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**Indian Institute of Technology, Madras**  
**Lecture - 20**  
**Analog Multipliers (continued)**

So, continuing with our lecture on transconductance type multiplier we saw that the Gilbert gain cell as depicted here has the two diodes connected in the following manner with currents  $I_1$  and  $I_2$  flowing through this.

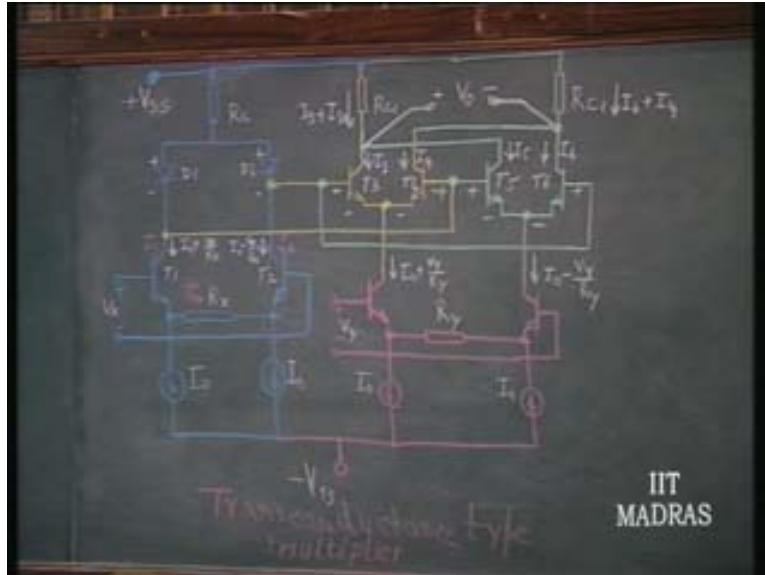
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We saw, if  $I_1$  and  $I_2$  are the currents and  $I_3$  and  $I_4$  are the currents of the transistors pair that is connected to the voltage resulting due to the two diodes then by applying the Gilbert's the translinear principle in the Gilbert's gain cell we can say that  $I_1$  into  $I_3$  so  $I_1$  into  $I_3$  is equal to  $I_4$  into  $I_2$  this is what we wrote in the last class also. Therefore,  $I_1$  by  $I_2$  is equal to  $I_4$  by  $I_3$  and then by using the ratio principle  $I_1$  minus  $I_2$  by  $I_1$  plus  $I_2$  therefore in turn is equal to  $I_4$  minus  $I_3$  by  $I_4$  plus  $I_3$ .

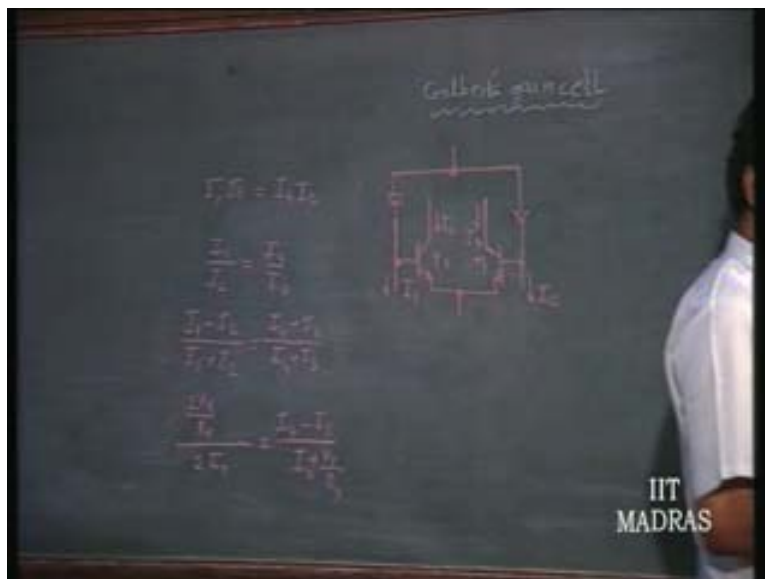
Now, here we recognize the fact that  $I_1$  and  $I_2$  in turn are differential input currents generated by using transconductor wherein a voltage  $V_X$  is applied and this voltage  $V_X$  will generate a current in this which is  $V_X$  by  $R_X$  neglecting these effects  $V_X$  by  $R_X$ .

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And therefore if this current being  $I_0$  and this current being  $I_0$  this will be  $I_0$  plus  $V_X$  by  $R_X$  and this will be  $I_0$  minus  $V_X$  by  $R_X$  where a differential current is generated. So if you replace this  $I_1$  and  $I_2$  by these two currents now we can see here that  $I_1$  minus  $I_2$  is nothing but, let us say this is our  $I_1$  and this is our  $I_2$ .

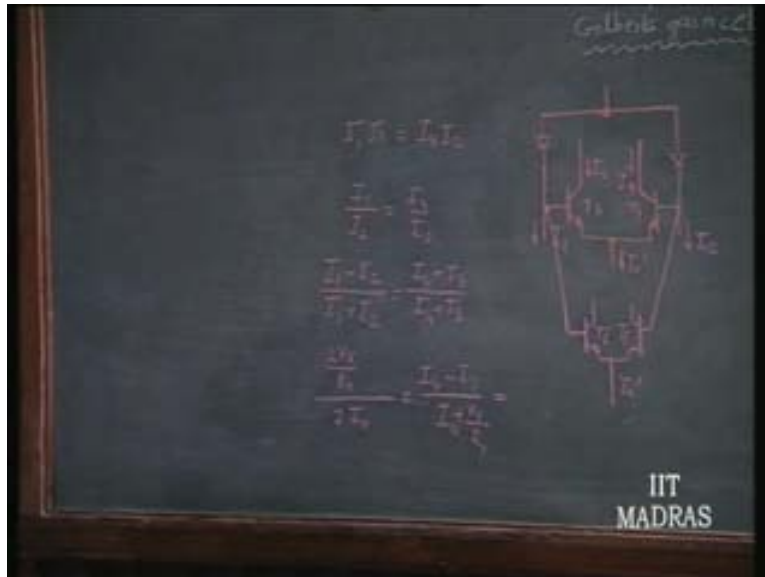
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So we will see that  $I_0$  plus  $V_X$  by  $R_X$  minus  $I_0$  minus  $V_X$  by  $R_X$  that is this is  $2V_X$  by  $R_X$  that is  $I_1$  minus  $I_2$  by  $I_1$  plus  $I_2$  this plus this is nothing but  $2I_0$ . This is now equal to the differential output current of this pair that is  $I_4$  minus  $I_3$  and divided by  $I_4$  plus  $I_3$  last time we called this current as  $I_0$  dash. So this is going to be  $I_0$  dash which is in this case  $I_0$  plus

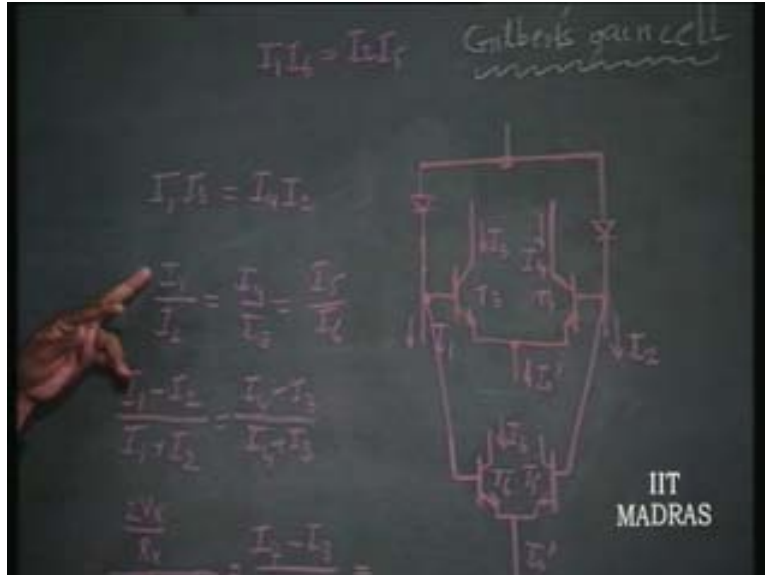
$V_Y$  by  $R_Y$  using the same principle of obtaining from a voltage a differential current by making a current of  $V_Y$  by  $R_Y$  flow through this. So once again the current in this is  $I_0$  plus  $V_Y$  by  $R_Y$  and this is  $I_0$  minus  $V_Y$  by  $R_Y$ . so we have now come to a conclusion that, by using a Gilbert gain cell one cell we can get this kind of a relationship.

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Similarly, we can show that this is also equal to by the same principle we have seen here that if you connect another transistor parallel to this call this  $T_3$  which is the transistor connected here and similarly the same base is connected to the other base of  $T_6$ ,  $T_3$  base and  $T_6$  bases are connected together and  $T_5$  and  $T_4$  bases are connected together. So it is just a parallel connection of another cell with another current flowing through this which is now going to be equal to, this was  $I_0$  prime earlier let us call it, this is going to be  $I_0$  double prime.

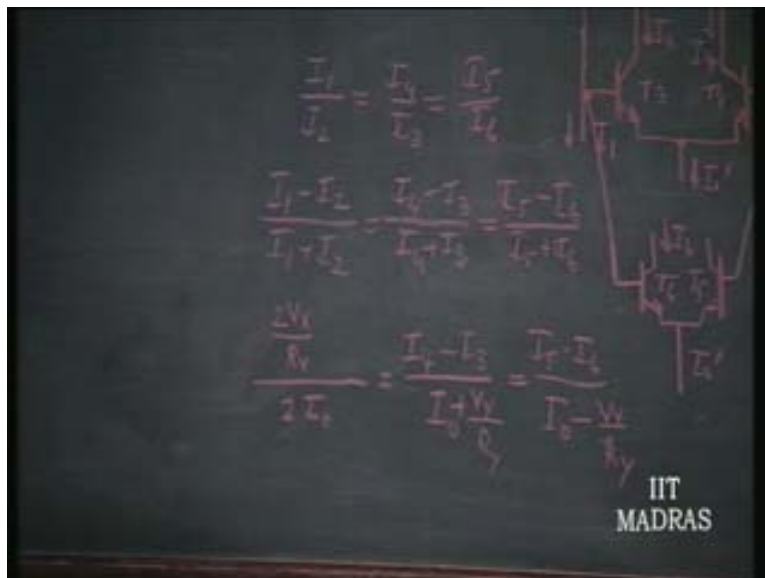
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So if this is the case by adopting a similar relationship here what will you get?

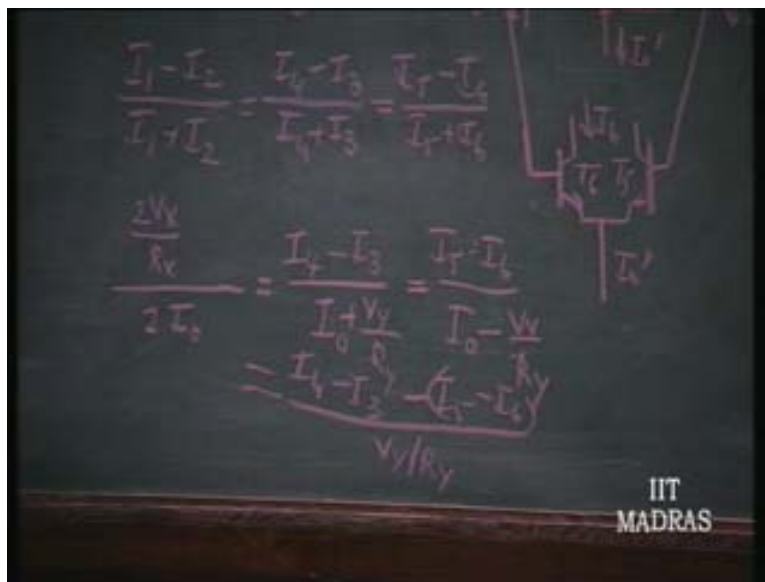
This voltage again remains the same so this diode current is  $I_1$  that into 6 that is the current in this so  $I_1$  into  $I_6$  is equal to  $I_2$  into  $I_5$  by using the same principle. We can keep on connecting any number of such cells in parallel. The ratio relationship remains exactly similar to what we had earlier which means again  $I_1$  by  $I_2$  from this is going to be equal to  $I_5$  by  $I_6$ . That means the ratio relationship  $I_1$  by  $I_2$  is equal to  $I_4$  by  $I_3$  is equal to  $I_5$  by  $I_6$  so on and so forth.

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Thus, adopting the same thing here we get  $I_4$  minus  $I_3$  by  $I_4$  plus  $I_3$  is same as  $I_5$  minus  $I_6$  by  $I_5$  plus  $I_6$  and we have obviously this differential output current of that transistor pair as  $I_5$  by  $I_6$  by  $I_5$  plus  $I_6$  is  $I_0$  double prime. And  $I_5$  plus  $I_6$  is  $I_0$  double prime which in turn is equal to in the diagram here  $I_0$  minus  $V_Y$  by  $R_Y$ . So it is obvious that if you want to get rid of  $I_0$  here the only thing you have to do is simply make this relationship equal to this plus this divided by this plus this. So,  $a$  by  $b$  is equal to  $c$  by  $d$  is same as  $a$  plus  $c$  by  $b$  plus  $d$ . so it tells us that in order to get rid of  $V_Y$  we can do this. If you want to get rid of  $I_0$  you have to subtract this from this so  $a$  minus  $b$  by  $b$  minus  $d$  which is also same as  $a$  plus  $c$  by  $b$  plus  $d$  any of these relationships. So, if you want to get rid of  $V_Y$  we have to add and if you want to retain  $V_Y$  and get rid of  $I_0$  we have to subtract.

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So we get  $I_4$  minus  $I_3$  minus  $I_5$  minus  $I_6$  by  $V_Y$  by  $R_Y$ . that means this relationship is same as this whole thing which is  $2V_X$  by  $R_X$  by  $2I_0$  equal to that is the final relationship  $I_4$  minus  $I_3$  minus  $I_5$  plus  $I_6$  if you remove the bracket so  $(I_4$  plus  $I_6$  minus bracket  $I_3$  plus  $I_5)$  rewriting the same numerator we get this as  $(2V_Y$  by  $R_Y)$ .

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$$\frac{\left(\frac{2V_x}{R_x}\right)}{2I_1} = \frac{(I_4 + I_6 - I_3 + I_5)}{(2V_x/R_y)}$$

$$\frac{(I_4 + I_6 - I_3 + I_5)}{2I_1} = \frac{4V_x V_y}{2R_x R_y I_0}$$

$$\frac{I_1}{I_2} = \frac{I_4}{I_3} = \frac{I_5}{I_6}$$

$$\frac{I_1 - I_2}{I_1 + I_2} = \frac{I_4 - I_3}{I_4 + I_3}$$

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Now what does it say?

Whatever is the summation of all these currents  $I_4$  plus  $I_6$  minus  $I_3$  plus  $I_5$  that thing is uniquely equal to  $4V_x V_y$  by  $R_x R_y$  into  $2I_0$ . You can see that we have to add current  $I_4$  to 6, how do you add currents? They are current sources, so  $I_4$  is added to  $I_6$  by just connecting it to a node. Those two currents get automatically added so this is  $I_4$  plus  $I_6$ . And here you have to add 3 to 5 that is  $I_3$  to  $I_5$  and then this node current is going to be  $I_3$  plus  $I_5$ . Then develop a voltage which is proportional to this difference. That means you can therefore take the output voltage between these two points, then what will be the output voltage?

If you take this as the output voltage then this point is nothing but  $V_{cc}$  minus  $I_3$  plus  $I_5$  into  $R_{Cy}$  minus  $V_{cc}$  minus  $I_6$  plus  $I_4$  into  $R_C$  which is nothing but  $I_6$  plus  $I_4$  into  $R_C$  minus  $I_3$  plus  $I_5$  into  $R_C$ .

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$$(I_1 + I_2 - I_3 - I_5) R_c = V_0$$

$$= \frac{2V_x V_y R_c}{2R_x R_y I_0}$$

$$= \frac{V_x V_y}{V_r} \cdot V_r = \frac{I_0 R_x R_y}{2R_c}$$

$$\frac{I_1}{I_2} = \frac{I_3}{I_5}$$

$$\frac{I_1 - I_2}{I_1 + I_2} = \frac{I_3 - I_5}{I_3 + I_5}$$

$$\frac{2V_x}{R_x} = \frac{I_1 - I_2}{I_1 + I_2}$$

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That means if you just multiply this by  $R_C$  this is going to be nothing but my  $V_0$ . You see how neatly the output voltage is going to depend purely on, obviously you can cancel this, this two times so we get the output as  $V_x V_y$  by  $V_R$  we will call it wherein  $V_R$  is equal to  $I_0$  into  $R_X$  into  $R_Y$  by  $2R_C$ .

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$$R_x = R_y = 2R_c = R$$

$$\frac{2V}{R} \cdot \frac{1}{2I} = \frac{(I_1 + I_2 - I_3 - I_5)}{(2V/R)}$$

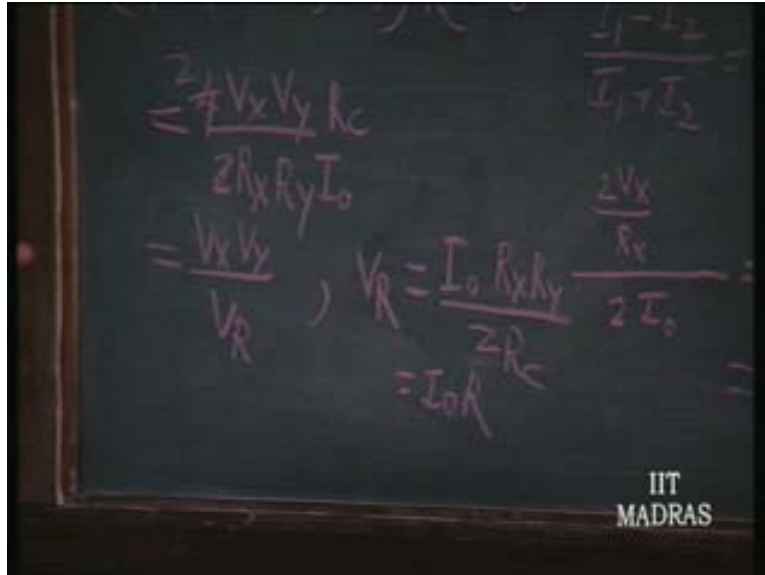
$$I_1 I_3 = I_2 I_5$$

$$\frac{I_1}{I_2} = \frac{I_3}{I_5}$$

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Let us say we make  $R_X$  is equal to  $R_Y$  is equal to  $R$  is equal to  $2R_C$  then this will be simply  $R_X$  is equal to  $R_Y$  is equal to  $2R_C$  is equal to  $R$  then this is going to be equal to  $I_0$  into  $R$ .

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This is an important design equation. What we should do is, we have to make  $V_R$  is equal to 10V. Therefore if  $I_0$  is given to you as 1 milliamperere  $I_0$  is the current source biasing current then automatically you know that  $R$  should be 10 kilo ohm because  $V_R$  has to be 10V. So, if  $I_0$  is given as 1 milliamperere  $R$  is coming out as 10 kilo ohms. That means  $R$  is 10 kilo ohms this is 10 kilo ohms  $R_X$  is equal to 10 kilo ohms  $R_Y$  is equal to 10 kilo ohm and  $R_C$  will be is equal to 5 kilo ohms.

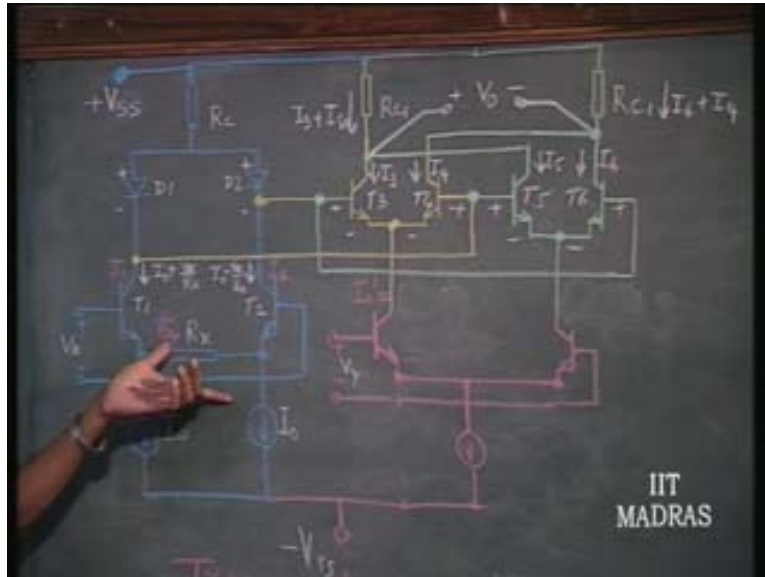
The design is over in the total sense because we have also fixed up the dynamic range now because we can see that if I now make  $I_0$  into  $R$  as 10V then this  $V_X$  can keep on changing until  $V_X$  by  $R_X$  is such that  $I_0$  minus  $V_X$  by  $R_X$  becomes equal to 0. That means the maximum  $V_X$  it can handle is automatically equal to  $I_0$  into  $R_X$  or  $I_0$  into  $R_Y$  and the maximum  $V_Y$  it can handle will be  $I_0$  into  $R_Y$  at which point of time one transistor will become off. So the moment I select this design here for a typical current  $I_0$  of 1 milliamperere I make  $R$  is equal to 10k,  $R_X$  is equal to 10k,  $R_Y$  is equal to 10k,  $R_C$  is equal to 5k and my  $V_X$   $V_Y$  limits also gets automatically fixed at minus 10 to plus 10V range. So the entire design is neatly over and this system is now ready for use as a precision multiplier.

Obviously these resistors  $R_X$ ,  $R_Y$  and  $R_C$ s have to be put externally if you want it to retain its precision property because all these dynamic ranges as well as the multiplication constant is determined by  $R_X$ ,  $R_Y$  and  $R_C$ . So this is one of the most popular IC multiplier available today. The only thing is it can be used as a precise multiplication up to about 1 MHz retaining the whole dynamic range of plus minus 10V and beyond that range there will be frequency limitation problems. That means up to about megahertz range this is very easily usable. But suppose in actual communication application we do not normally have signal input which is of the order of 10V there is no question of that and it is going to be of the order of micro volts or millivolts. So there is no point in improving the dynamic range. Therefore for communication application we can simply say that there is



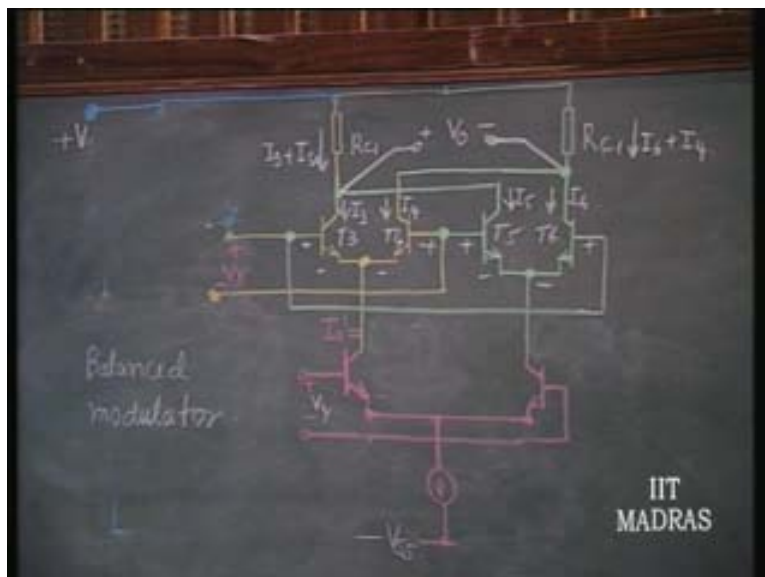
no need to put this that means this can be shorted and now there can be a single current source but the only thing is it will not be  $V_Y$  by  $R_Y$  now but it will be  $V_Y$  into  $g_m$  that is  $1$  by  $2R_E$  of the transistor.

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This is the modification for a communication IC. Therefore here also I do not have to do this modification of converting voltage into differential current like this, I get rid of the whole stuff because in any case for small signal the differential gain itself is getting multiplied by  $g_m$  which is directly proportional to the current, so that is a multiplier.

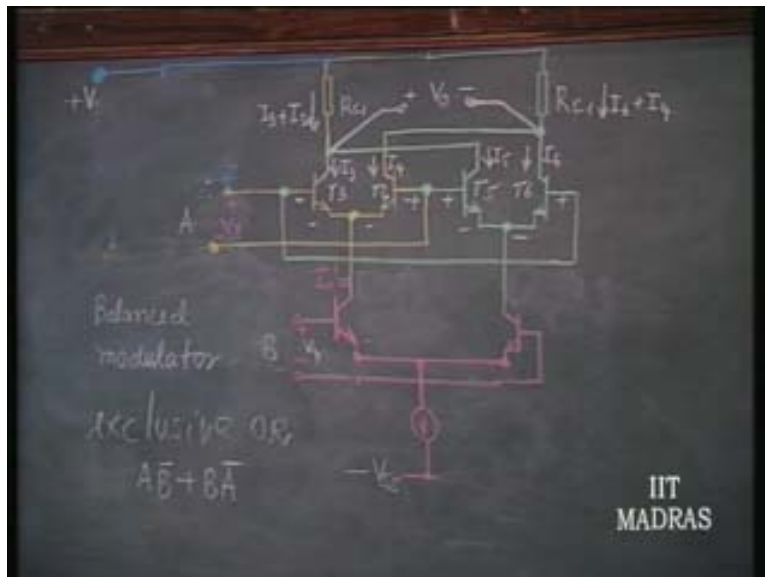
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So this circuit is called as balanced modulators by the communication engineers. This is separately available as another IC, so we have got rid of those linear transconductors which we had put to convert  $V_X$  and  $V_Y$  into differential currents and the circuit becomes pretty simple. And this is very commonly used by communication engineers obviously for multiplication purpose which is normally termed by communication terminology as balanced modulator ICs. This can be used up to hundreds of mega hertz. Therefore it is very popular for the purpose of fixing. All these operations are multiplication operations namely mixing, modulation, this is also called balanced modulator cum demodulator because the operation of modulation or demodulation both operations mean multiplication, balanced modulation and balanced demodulation.

So this particular IC is a very popular digital IC and the operation that is performed, if this is high or low and this is again high or low. That means now we are not talking of analog signal, this goes high or low and this also goes high or low independently. What is the output it is going to give? This circuit, when this goes high this goes high you find out the output and when this goes high and this goes low and you find out the output so do this and find out the truth table for this and you will see that this is called a gate Exclusive OR.

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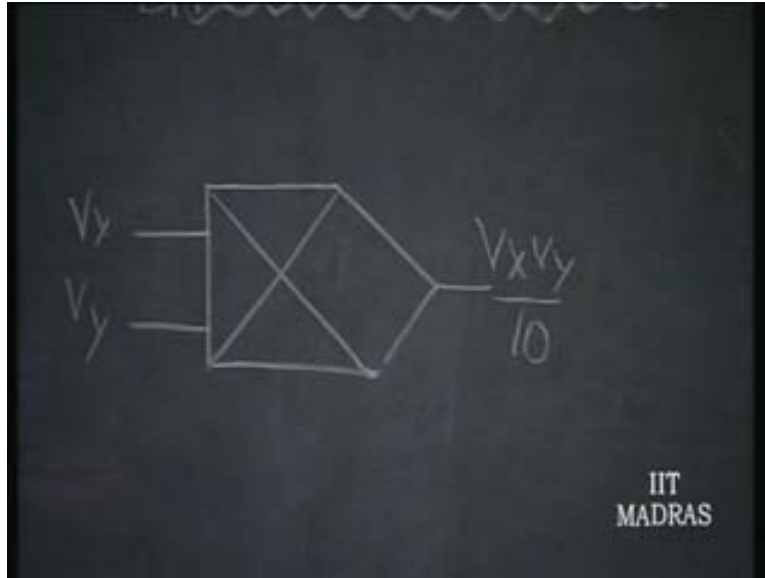


If there are two inputs a and b what is the exclusive OR operation?

It is  $a\bar{b}$  plus  $b\bar{a}$ . So it is called exclusive OR. So if you look at a circuit you should be able to identify it straight away in ECL Emitter Coupler Logic. The fastest exclusive OR gate that you can think of for the purpose of this logic operation which is a  $\bar{a}b$  when this is a and that is b. **You can find out the output corresponding to both high both low one high and one low etc and see how the output looks like.** Output can in turn be limited by putting diodes across the output voltage so that the output does not swing as determined by the current but is limited to plus minus  $V_{\gamma}$ . This circuit is also used as phase detector. Let us first understand the important communication application of this

multiplier. Let us now look into the application to understand the difference between balance modulator, phase detector and precision multiplier.

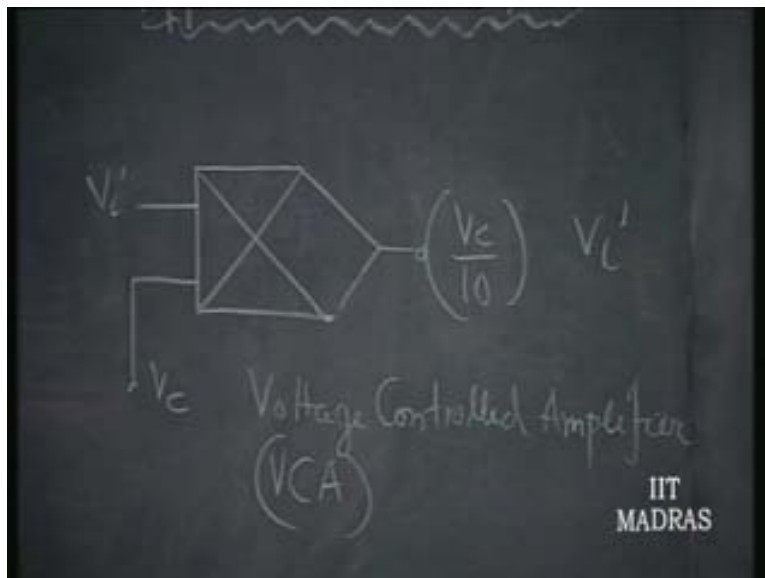
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Application of multipliers:

We will assume that is a block obtained by anyone of those methods but it is a four quadrant multiplier as given here  $V_x$  and  $V_y$  being input and the output is  $V_x V_y$  by 10 and the dynamic range being plus minus 10V.

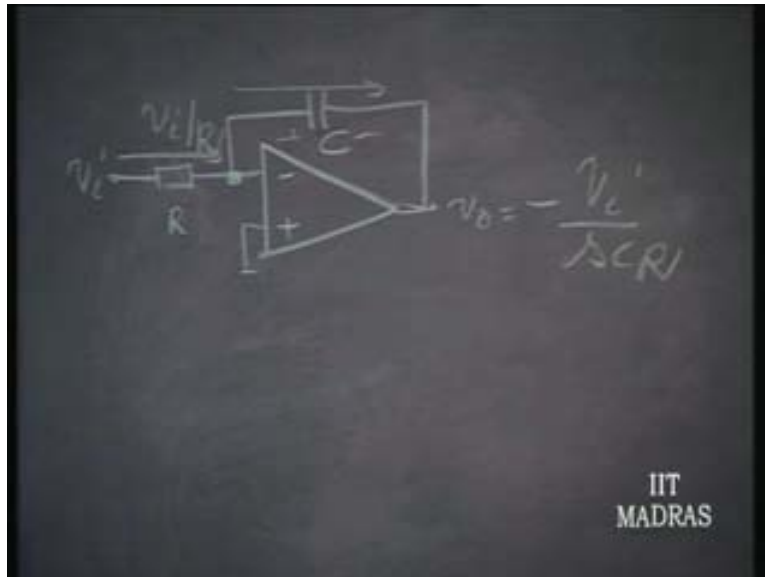
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Let us try to use this ideal multiplier for a variety of applications. The first application wherein it is used in large numbers is where  $V_X$  is going to be equal to the input voltage and  $V_c$  is going to be equal to the control voltage then the output is just used as a multiplier straight away without any additional paraphernalia being added here so  $V_c$  by 10 into  $V_i$ . So this is what is called as voltage controlled amplifier but it does not amplify in terms of voltage gain but in terms of power gain is what it really means when we say amplifier so still it can be called an amplifier. Therefore this is called voltage controlled amplifier.

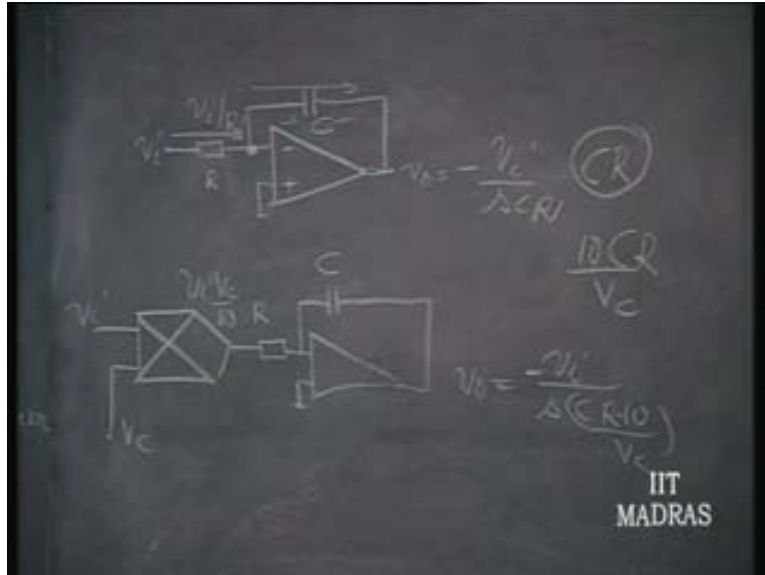
In turn we can say that if you can design a linear voltage controlled amplifier it is a candidate for multiplier. The gain can be controlled from 0 to 1 it can change signs. That means it is acting both as inverting and non inverting amplifier depending upon the sign here. You have control over the magnitude of gain as well as the sign which is an important thing. So this voltage controlled amplifier is commonly used in what are called music synthesizers and things like that. In most of the music synthesizers the special effect generators etc this voltage controlled amplifier is used in large numbers.

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Voltage controlled amplifiers can in turn be used, this is an op amp integrator when this is  $V_i$  the current in this is  $V_i$  by  $R$  it will flow through this and develop a voltage which is  $V_0$  is equal to minus  $V_i$  by  $sCR$  so it is an integrator. An integrator is one of the most important basic building blocks in analog signal processing particularly in what is called filtering. Therefore an integrator is an important building block.

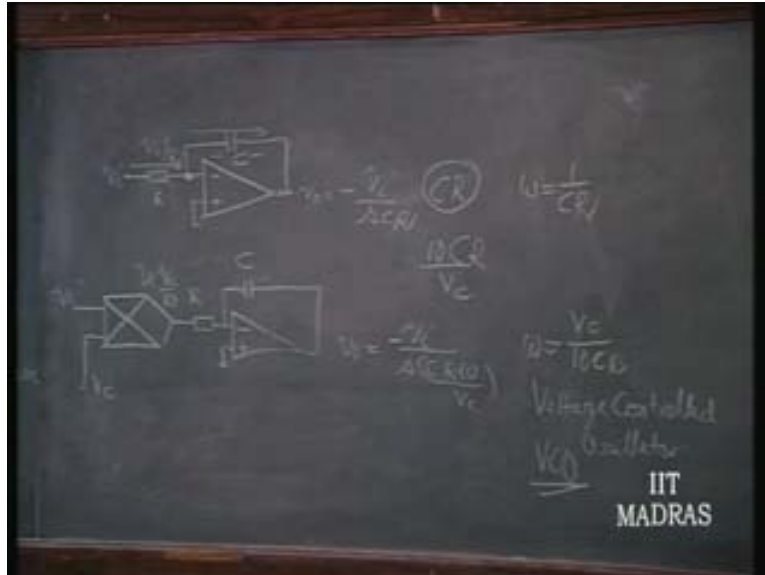
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Therefore now if I want to convert this integrator to become voltage control I merely have to put this before that. Now you see that it becomes an integrator which is voltage controlled because originally we were feeding  $V_i$  here and now we are feeding  $V_c$  here so this becomes  $V_i$  into  $V_c$  by 10. So, wherever originally we had  $V_i$  we simply put  $V_c$  by 10 and we get the transfer function for this that is minus  $V_c$  by  $10sCR$  into  $10$  by  $V_c$  because it is  $V_c$  by  $10$  wherever  $V_i$  is there we put  $V_c$  by  $10$ . This means the time constant  $CR$  is merely changed to  $CR$  into  $10$  by  $V_c$ . That means in any circuit that was earlier using this integrator where the time constant was  $CR$  I simply replaced it by this indicator and replace the time constant  $CR$  by  $10CR$  by  $V_c$ .

I get the performance factors of that particular circuit. Now let us say I use this integrator in what we are calling as a double integrator oscillator that we consider now. A double integrator oscillator is something in which two integrators are connected together and then an inverter is connected and if you now close the loop it becomes a double integrator oscillator or a harmonic oscillator. It just simulates a second order differential equation  $d^2 V_0$  by  $dt^2$  plus  $KV_0$  is equal to 0.

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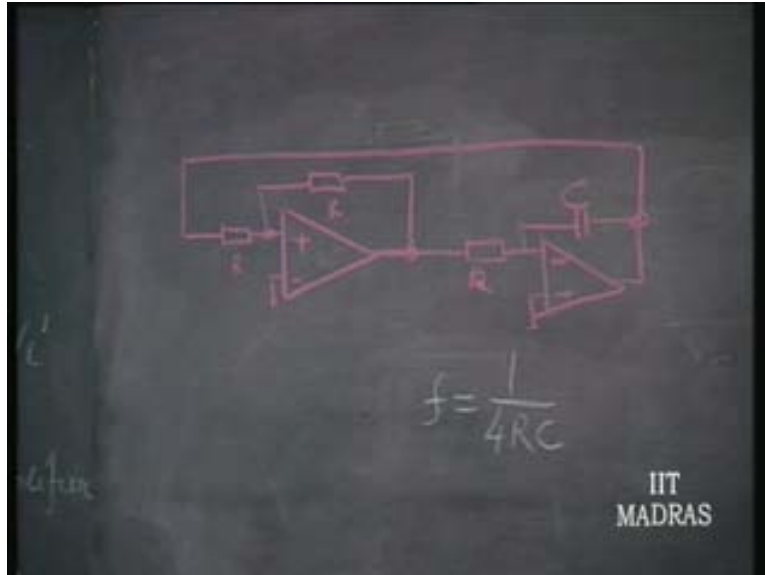
Therefore that is a double integrator oscillator and if you replace it by this the frequency of oscillation of the double integrator oscillator is going to be  $\omega$  is equal to  $1/CR$  whereas in this case it will be  $\omega$  is equal to  $V_c/10CR$ . That means it becomes a voltage controlled oscillator. It becomes a linear voltage controlled oscillator or VCO. The frequency of oscillation can be linearly controlled by varying  $V_c$ .

What does this circuit mean in terms of communication?

You are having a FM generator. That means now if I have a carrier to be frequency modulated I simply apply the frequency the modulating voltage in series with some constant voltage.  $V_c$  will be some constant which will determine the carrier plus  $V_m \sin \omega_m t$  then I can generate FM. So, voltage controlled oscillators are used for FM generators.. Communication engineers therefore build FM generators in order to build voltage controlled oscillators.

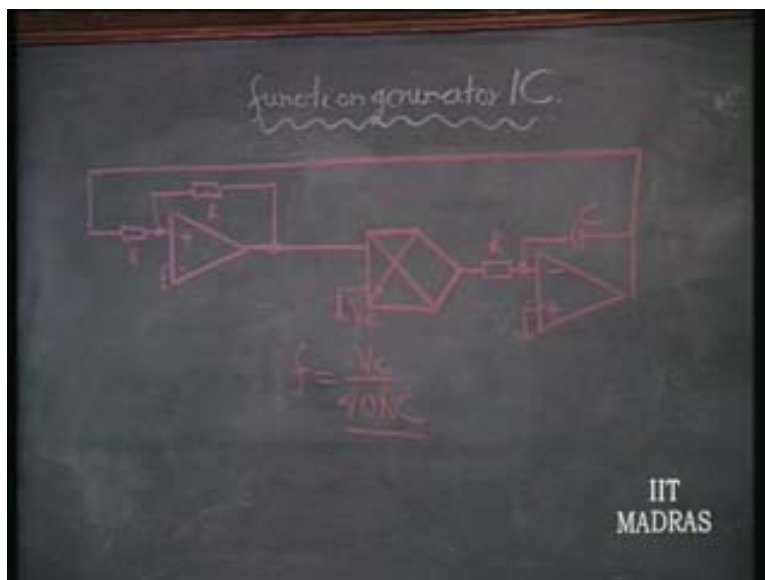
Now if the same block is used in a filter configuration which is called KHN filter, again it is nothing but the double integrator filter, the center frequency is going to be  $\omega$  is equal to  $1/CR$  if it is band pass output and in the voltage controlled version you will have that being controlled by  $V_c$  linearly. So you can now design a voltage controlled filter. All these things are simple consequences of replacing the integrator by the voltage controlled integrator. So you can see how these variety of circuits get generated simply by obtaining this voltage controlled action of the multiplier.

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For example, here is something again used as an IC, this circuit is a very popular circuit. This circuit is called the function generator circuit. This gives you a square wave and this gives you a triangular wave. You have frequency omega or frequency  $f$  is equal to  $1$  by  $4RC$ . That means now I want to convert it into a voltage controlled oscillator. Now it is enough if I just replace this one integrator by the multiplier. The advantage of this voltage controlled oscillator over the previous one is simply that it uses only one multiplier as against two multipliers needed in the other circuit.

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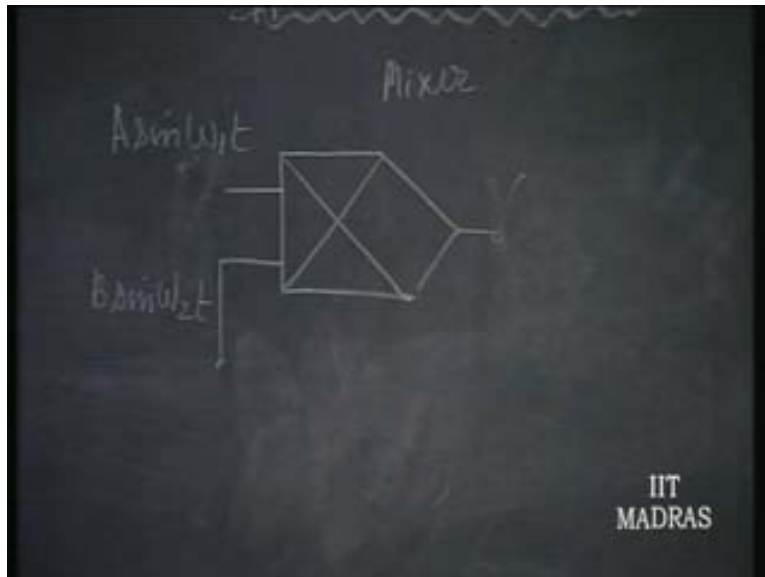
So this is presently being used in most of the ICs for the purpose of voltage controlled oscillator with facility for linear modulation of the frequency. So what happens now?

The frequency of oscillation becomes, RC has to be replaced by  $10 RC$  by  $V$ . This now becomes  $f$  is equal to  $V_c$  by  $40RC$ . This whole thing is called a function generator IC with the multiplier, a Schmitt trigger and an integrator. It only gives you a triangle not a sine wave but for most of the communication applications we can use this itself as the FM output it does not matter. Hence it is perfectly linear.

Later on we will discuss this entire thing as a voltage controlled oscillator separately using transistors. These use op amps. Obviously the frequency up to which you can use this is highly limited not by the principle but by the op amp you are using. The op amp, slew rate etc will limit the performance of this IC. These are large signal oscillators so slew rate comes into picture. Therefore instead of realizing this using operational amplifiers and multipliers you can do this simply by using transistors.

Later on let us look into how transistor voltage controlled oscillator circuits can be generated which also uses the same principle of a Schmitt trigger a voltage controlled conductor charging a capacitor. But function generator IC is an important application of multiplier, the FM generator.

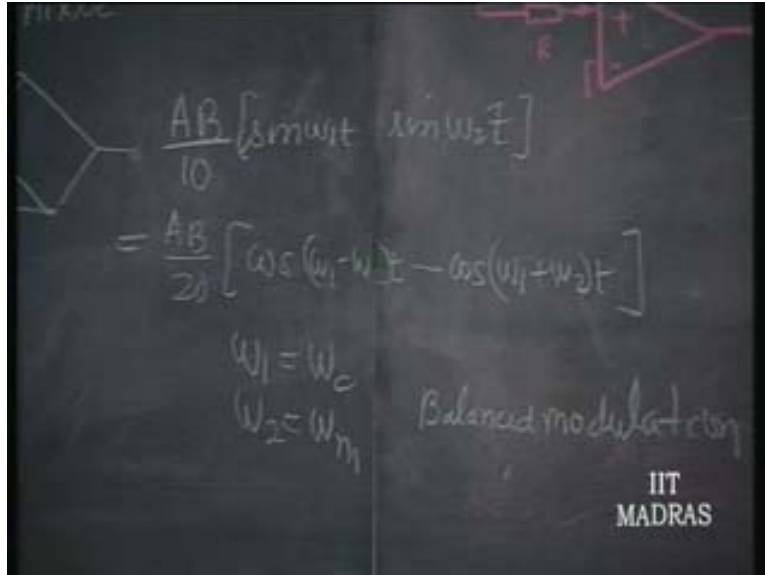
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Now, if I make one input equal to  $A \sin \omega_1 t$  and the other input as  $B \sin \omega_2 t$ , now the multiplier is still being used as multiplier but inputs change. In the earlier situation we had one as conventional input of any frequency but the other as DC then it became a voltage controlled amplifier. Here both the inputs are AC of two different frequencies. This operation of multiplication in communication terminology is called mixing. It is just mixing of two frequencies so it is called a mixer.



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The image shows a chalkboard with handwritten mathematical equations and text. The equations are:

$$\frac{AB}{10} [\sin \omega_1 t \sin \omega_2 t]$$
$$= \frac{AB}{20} [\cos(\omega_1 - \omega_2)t - \cos(\omega_1 + \omega_2)t]$$

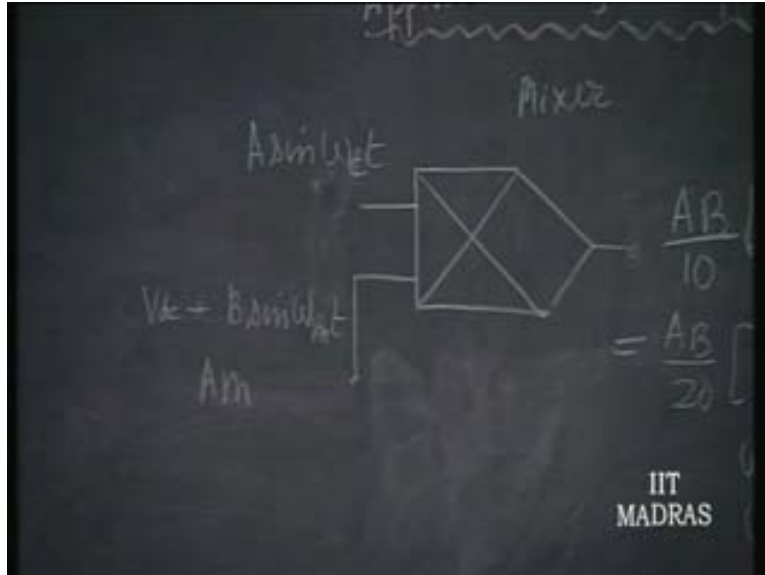
Below the equations, it is noted that  $\omega_1 = \omega_c$  and  $\omega_2 = \omega_m$ . The text "Balanced modulation" is written to the right of the equations. In the bottom right corner, the logo for "IIT MADRAS" is visible.

And obviously the output will be  $\frac{AB}{10} \sin \omega_1 t \sin \omega_2 t$  is equal to  $\frac{AB}{20} [\cos(\omega_1 - \omega_2)t - \cos(\omega_1 + \omega_2)t]$  where these are called side bands. If  $\omega_1$  is the carrier and  $\omega_2$  is the modulating frequency then this application is called balanced modulation. These are the two sidebands  $\omega_c - \omega_m$  and  $\omega_c + \omega_m$ . So these are the two sidebands that you are getting. This is not the amplitude modulation but this is balanced modulation, the carrier is absent.

If you introduce the carrier, how do you introduce a carrier?

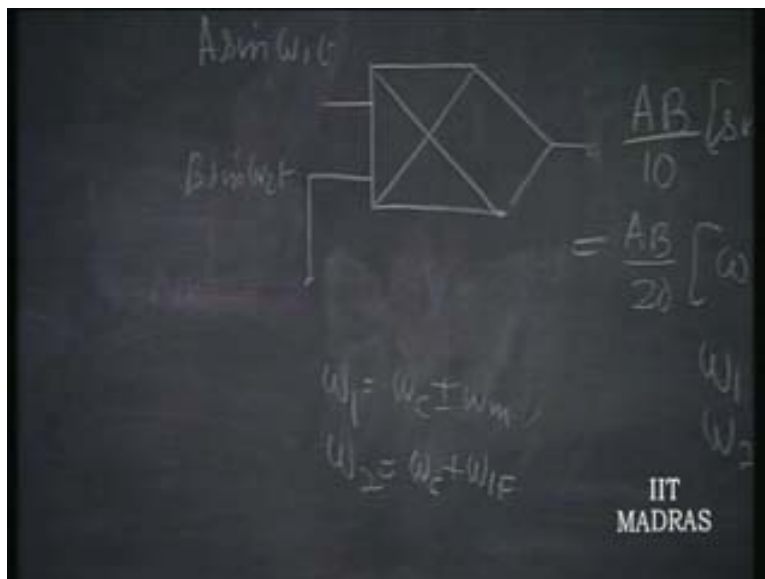
The only thing is, you should not multiply, this is  $\omega_c t$  but this is  $\omega_m t$  plus  $A_{dc}$  then automatically carrier will come or it will be having feed through components. That means actually for producing amplitude modulation you do not need an ideal multiplier but you need a non ideal multiplier. So, amplitude modulation means  $V_{dc}$  plus this and this is  $\omega_m$  and this is  $\omega_c$  that is the amplitude modulated output.

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So, by nearly changing the input I am getting different applications. For AM generation the input will be this and for balanced modulation the input will be this. If on the other hand,  $\omega_{a1}$  is  $\omega_c$  plus or minus  $\omega_m$  and  $\omega_{a2}$  is  $\omega_c$  plus  $\omega_{IF}$  what happens?

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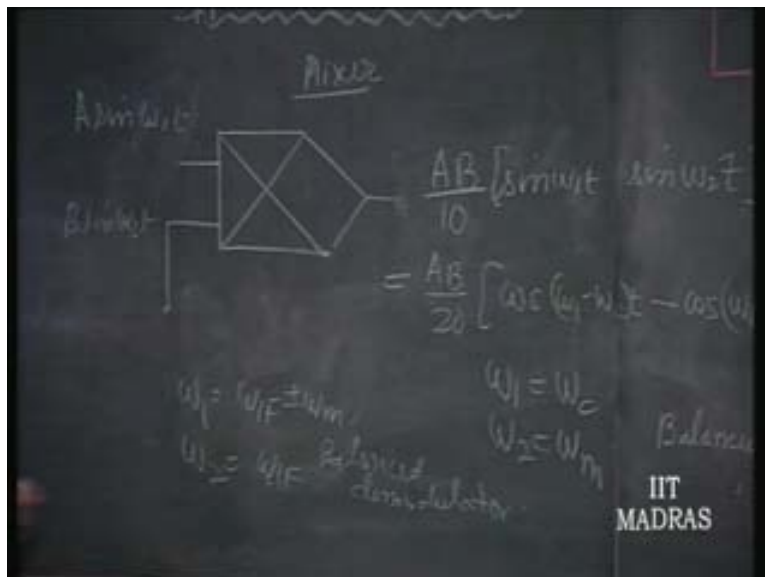
In the difference component you will get  $\omega_c$  going and it will be  $\omega_{IF}$  plus or minus  $\omega_m$ . There will be of course two  $\omega_c$  components plus  $\omega_{IF}$  plus or minus  $\omega_m$  component the other one. That can be got rid of by using a band pass filter which is called IF filter, this is called IF filter. That means at that point of time this is

taking the role of a true mixer. And this generated by what is called local oscillator that is the basic principle of super heterodyne receiver. The radio receiver or for that matter even television receiver. That means you are having an IF.

For example, it is like different people carrying the same luggage may be from the railway station to the taxi stand a porter is carrying thereafter once it comes to the house you have to carry it there is no other go, so that is the thing. Therefore basically we can design these local things to operate at a fixed frequency. Therefore we can design better amplifiers at fixed frequency than at variable frequency. That means RF amplifiers are more difficult to design than IF amplifiers. that is the reason we go in for amplification at a fixed frequency and transfer the entire responsibility of amplification or what is called sensitivity of the radio receiver or a television receiver to what is called an IF stage. That design is done at fixed frequency and there is no problem whereas most of the carriers will be coming at different frequencies. These are selected and the entire information is transferred, it is like merely acting as a porter carrying the information but a common porter.

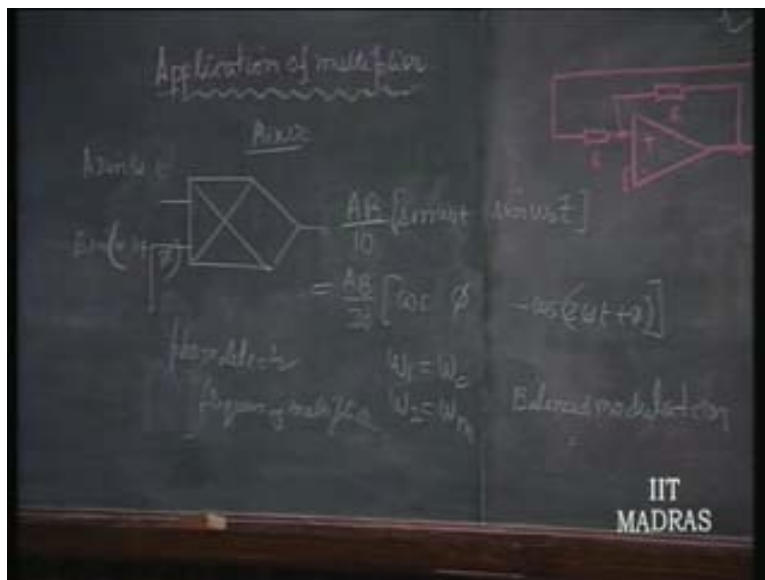
As soon as I come to my house instead of me carrying I can always call may be call my servant who is going to carry this and any number of times I come the same person will be carrying whereas from the railway station to the house it will be different persons for carrying. Then there might be some trouble in dealing with such people whereas I have trained my servant and have given him enough rewards so there is no likelihood of any complaint. The same thing is with the design of IF stages. It has been optimized and therefore it is going to do the business without much of a problem. This particular thing is going to be then called mixer. So we have to now see how inputs can change the application of the particular multiplier. The name also is different.

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So  $\omega_1$  is  $\omega_{IF}$  plus or minus  $\omega_m$  and  $\omega_2$  is  $\omega_{IF}$  which is easily generated locally and the output will be  $2\omega_{IF}$  plus or minus  $\omega_m$  but it can be got rid of by using a low pass filter and the only component you get will be  $\omega_m$  component which is the modulating frequency component. That is nothing but demodulation of the balanced modulator. This demodulation technique is not adopted in the conventional receivers simply because it requires the generation of  $\omega_{IF}$ . It can be done simpler if you adopt amplitude modulation that is adopted only for the fact that it can be simply detected by chopping of one portion and doing envelope detection using a simple diode. Therefore balanced demodulation becomes imperative if balanced modulation is there, carrier is not there and you are able generate equal stability carrier at the receiving point which is a tough affair. So this is the balanced demodulator. Now, if  $\omega_1$  is  $\omega$  and  $\omega_2$  is also  $\omega$  but then if they are two same frequencies they can be differing in phase.

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Then what do you get as the output here? This will be  $\phi$  and then that will be  $2\omega$ , plus  $\phi$ . Once again there are two applications. So  $AB$  by  $20 \cos \phi$  that means it develops a DC voltage which is dependent upon phase shift. So it is defined now as a phase detector. It is not a linear phase detector but it is a cosine FM. Again you can see you have to put a low pass filter in order to get rid of  $2\omega$  component. You put a high pass filter you get rid of the DC and you will get frequency doubling. The same thing can be used as a phase detector or frequency multiplier depending upon whether you connect a low pass filter or a high pass filter.

Now again, if  $A$  is a voltage which is proportional to line voltage and  $B$  is a voltage proportional to line current then it is nothing but power, it is like an electronic watt meter. So it is again used for such applications. I mean you have to use a current transformer and a voltage transformer respectively in order to generate these input voltages and feed it to this multiplier at the proper level of voltage. Then you can get power factor or power etc.

So you can use your 2 watt meter method and instead of 2 watt meter method of power measurement in three phase circuit you can use two multiplier method of power measurement, active or reactive whatever it is. So a variety of such applications are possible with the help of the versatile tool which is the multiplier. You can multiply frequency, if you can multiply frequency you can generate harmonics,

if you can generate harmonics you can synthesize waveform, you can synthesize any periodic waveform in terms of  $t$  square,  $t$  cube and all these things can be done. So a nonlinear functional block like a multiplier is used for function synthesis. Any non linear function can be generated to any accuracy you want if you are prepared to use any number of multipliers. These are the powerful applications of this multiplier. In the next class we will continue further with other varieties of multiplier that are available in integrated circuit form.