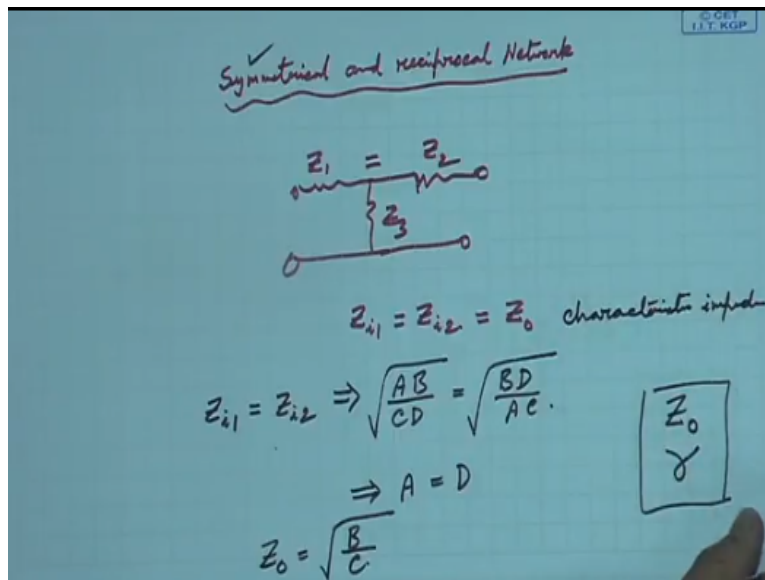


Design Principles of RF and Microwave Filters and Amplifiers
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Module No # 1
Lecture No # 03
Symmetrical lossless network description for filter design

Welcome to this third lecture now, the same 2 port reciprocal network we enforce symmetry in the network. Generally filters, most of the times we design symmetrical filters the 2 port network is symmetrical. What do you mean by symmetrical? When I said this 2 port that I may say this is Z_1 Z_2 Z_3 . So in a T section as I already said that any 2 port network whatever may be the interconnection can be always represented like this now this Z_1 Z_2 different that is a symmetric network.

(Refer Slide Time: 00:39)



if I enforce symmetry then $Z_1 = Z_2$ so what happens to Z_{i1} my image impedance Z_{i1} and you see the expressions of Z_{i1} and Z_{i2} already derived the moment Z_1 and Z_2 are same they becomes equal. So for a symmetrical network Z_1 Z_{i2} are equal that we generally call Z_0 . What is Z_0 a very very important concept called characteristic impedance.

So for a symmetrical network I have correct only one characteristic impedance for a non-symmetrical reciprocal network I have two image impedance but here it is one. So, what is the

concept of characteristic impedance? that I terminate with characteristic impedance then the source with characteristic impedance Z_0 will see that input impedance is also Z_0 , so it can deliver maximum power okay. But now so we see that we have characteristic impedance let us see what is his value in terms of ABCD parameters.

We already seen that when we enforce $Z_1 = Z_2$ that mean symmetry Z_1 becomes Z_2 so that implies that we have already earlier derived $Z_{in} = AB + CD$ and this is $BD + AC$ so that becomes $A = D$. You all know into port parameters transmission parameters if symmetrical network is equal to D it comes from here so what happens to value of Z_0 it is simply B by C .

And so now you see and that for a symmetrical reciprocal network I have I can characterize the network by only 2 parameters 1 is that this Z_0 and another we have already seen it is γ . So Z_0 and γ are sufficient to characterize the network so by Z_0 we will understand what will be our impedance level if there is an impedance mismatch etc and with γ will enforce whether the frequencies passed or not whether attenuated or not.

So let now express the ABCD parameters we are so familiar with in terms of Z_0 γ that means from now on will specify only Z_0 and γ for filter etc specification so we should have a mapping between ABCD etc. So you can easily do that express ABCD in terms of Z_0 and γ because now our description will be in terms of that so you can start from that $AD - BC = 1$ and also $A = D$.

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A, B, C, D in terms of γ

$$\begin{cases} AD - BC = 1 \\ A = D \end{cases}$$

$$BC = \sqrt{A^2 - 1}$$

$$\frac{B}{C} = z_0 \quad z_0 = \sqrt{\frac{B}{C}}$$

$$C = \frac{\sqrt{A^2 - 1}}{z_0}$$

$$B = z_0 \sqrt{A^2 - 1}$$

$$e^\gamma = \sqrt{AD} + \sqrt{BC} = A + \sqrt{A^2 - 1}$$

$$A = \frac{e^{2\gamma} + 1}{2e^\gamma} = \frac{e^\gamma + e^{-\gamma}}{2} = \cosh \gamma$$

$$B = z_0 \sinh \gamma$$

$$C = \frac{1}{z_0} \sinh \gamma$$

$$D = \cosh \gamma$$

So with this you can now find out that let us decide at what will happen to them if we put this $A = D$ we get $BC = \text{root over square} - 1$. And we know that B by C is Z_0 square you see this one we have found as B by C so from this you can find out the value of C is $\text{root over square} - 1$ by Z_0 and B is $Z_0 \text{ root over square} - 1$ but and what is A to the power γ earlier we have seen it is $AD + BC$. Now in terms of this I can write it as $A + \text{root over square} - 1$ so this equation I can solve so A will be expressed in terms of A to the power γ .

So by doing that I can see that A becomes A to the power $2\gamma + 1$ by $2A$ to the power γ that is nothing but A to the power $\gamma + A$ to the power $-\gamma$ by 2 which is nothing but $\text{COS hyperbolic } \gamma$. So A is nothing but $\text{COS hyperbolic } \gamma$ now put this B is already expressed in terms of Z_0 and this Z_0 value.

You know let me write $\text{root over } B$ by C so you can call for and see that B will be equal to $Z_0 \text{ SIN } \gamma$ please all of you do this exercises I am just writing you cannot memorize all these things but I am telling you the way so please do this manipulation it requires a bit familiarity with hyperbolic functions and $C = 1$ by $Z_0 \text{ SIN hyperbolic } \gamma$ and $D = \text{COS hyperbolic } \gamma$.

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Handwritten equations on a grid background:

$$A = \cosh \gamma$$

$$B = Z_0 \sinh \gamma$$

$$C = \frac{1}{Z_0} \sinh \gamma$$

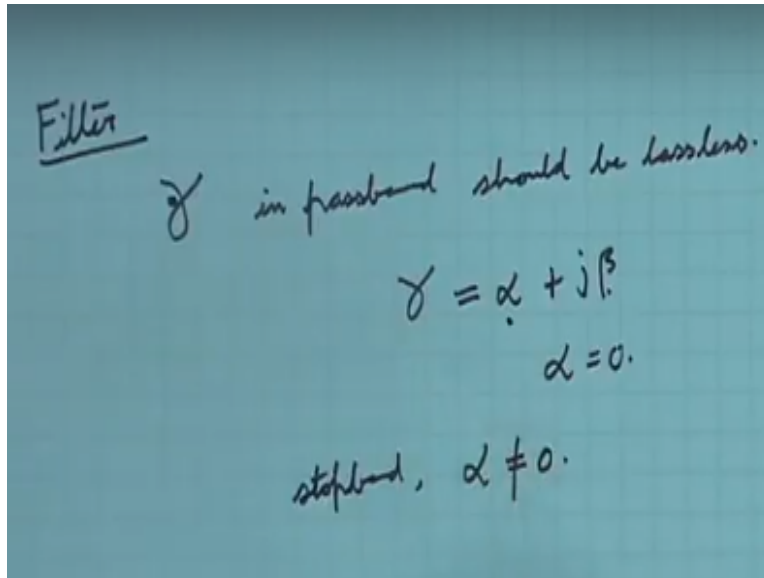
$$D = \cosh \gamma$$

Parameters: γ, Z_0

So you see that this expression that means now I can clearly write that $A = \text{COS hyperbolic gamma}$ $B = Z_0 \text{ SIN hyperbolic gamma}$ $C = \frac{1}{Z_0} \text{ SIN hyperbolic gamma}$ and $D = \text{COS hyperbolic gamma}$ so you see that actually here from also you can see that ABCD parameter was not needed because $A = D$ for a symmetrical network.

And this is the description that any ABCD parameters I can express in terms of 2 quantities is gamma and Z_0 . So that is the beauty you know that we are in the right path what we will do now will put the filter condition. So for a filter as I said what is the filter that it will allow some bands to pass it will stop all other bands. So we demand that now 2 port network is in terms of gamma and Z_0 what is our demand that if it is to be a filter so gamma in pass band should be lossless.

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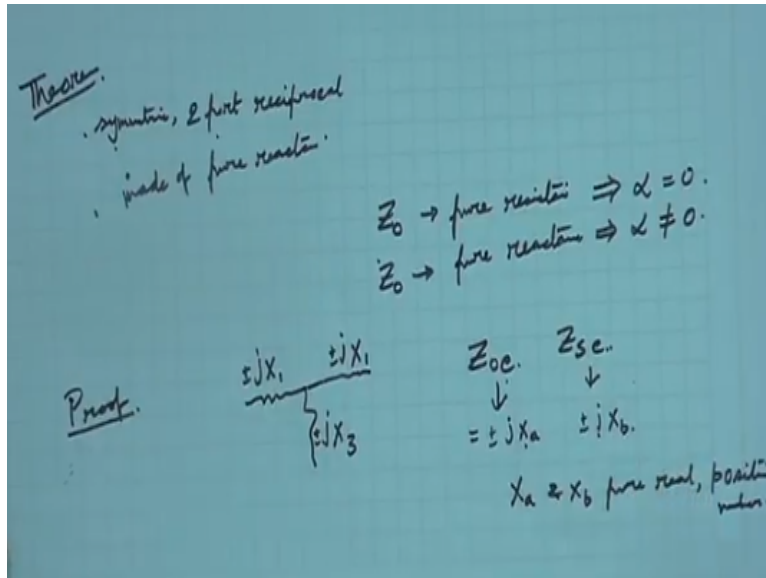


We want that gamma the propagation constant should be lossless then only the circuit won't attenuate power also not store power it will be able to deliver it to the port 2. So but you know that gamma is a complex quantity as I said that how we define that input volte ampere and output volte ampere their ratio that we call it A to the power 2 gamma.

So gamma is that now you are also familiar that generally this complex quantity we write as 2 part alpha + J beta, alpha is attenuation constant beta is the phase constant. For any electromagnetic wave this is true so when we say alpha in the pass band should be lossless. What we basically say that alpha should be 0 over the whole pass band that there should not be any attenuation. And what happens in stop band we want that it should be lossy ,so atleast alpha should not be = to 0.

Ok and we have already decided that to have this pass band alpha = 0 we will use only reactive elements. Now there is a theorem which will now see the theorem says that for a symmetric 2 port reciprocal network of pure reactances. If the characteristic impedance is a pure resistance that innovation constant is zero. If the characteristic impedance is pure reactance their attenuation constant is non-zero. So for this symmetric 2 port reciprocal this is a theorem of Everett symmetric 2 port reciprocal network made of pure reactances.

(Refer Slide Time: 10:33)



The theorem says that if Z_0 is pure resistance then $\alpha = 0$ if Z_0 is pure reactance α is non-zero let us prove this theorem. So will now prove this theorem since the network is made of pure reactances can I say that network these are all instead of Z they are jX it is a symmetric network jX and let us say X_3 instead of Z_1, Z_2, Z_3 now this may be plus minus whatever if this is the case then can I say that Z_{OC} and Z_{SC} . Which we already saw they are also pure reactances. We have seen that $Z_0 = \frac{Z_{OC} Z_{SC}}{Z_{OC} + Z_{SC}}$ into Z_{SC} open circuit and short circuit cases.

So they will be because we have seen their various combination of these they will be also pure reactances. So let us consider that Z_{OC} , let us call it either $+$ or $-$ the pure reactance this is let us call $+$ or $- jX_B$. That means X_A and X_B pure real positive numbers you agree that it should be. Because since I have incorporated jX_A and X_B are positive real numbers now let us see there can be several cases.

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Case	Z_{oc}	Z_{sc}	$Z_0^2 = Z_{oc} Z_{sc}$	$\tanh^2 \gamma = \frac{Z_{sc}}{Z_{oc}}$
I	$+jX_a$	$-jX_b$	$X_a X_b$	$-X_b/X_a$
II	$-jX_a$	$+jX_b$	$X_a X_b$	$-X_b/X_a$
III	$+jX_a$	$+jX_b$	$-X_a X_b$	X_b/X_a
IV	$-jX_a$	$-jX_b$	$-X_a X_b$	X_b/X_a

For case I & II, Z_0 is real and $\tanh \gamma$ is imaginary
 For case III & IV, Z_0 is imaginary and $\tanh \gamma$ is real.

Case 1, 2 am proving the theorem, is important for filter design now let me say what is ZOC what is ZSC then let me we know $Z_0^2 = ZOC, ZSC$ and we now 10 the square is ZSC by ZOC so we will make table.

So case 1 is let us say that this ZOC and ZSC they are of same sign so it is $+jX_a$ it is $-jX_b$ then what will be Z_0^2 the it will be simply $X_a X_b$ and what will be these ZSC by ZOC it will be $-X_b$ by X_a case 2 this is $-jX_a$ this $+jX_b$ this will be $X_a X_b$ this will be $-X_b/X_a$ this is $+jX_a + jX_b$ this is $-X_a X_b$ this is X_b by X_a this is $-jX_a - jX_b - X_a X_b$ X_b by X_a .

So you see that time I said that I have some possibilities so I have taken all permutations of these. From here you can see that case 1 and 2, Z_0 is real that means Z_0 it is a characteristic impedance it is real means this is a pure resistance and TAN hyperbolic is imaginary. Because TAN this square is $-$ so TAN hyperbolic will be an imaginary number. And in case 3 and 4, I have marked that by rate you see Z_0 is imaginary and TAN hyperbolic is real. So let me write this thing that in case 1 and 2 Z_0 is real and TAN hyperbolic is imaginary.

In our power cases 3 and 4, Z_0 is imaginary and TAN hyperbolic gamma is real. Still I have not proved the theorem because theorem said that if Z_1 is real, alpha will be 0, if Z_1 is imaginary there will be non-zero alpha. I have not reached there. So I need to do that to prove the theorem

so part 1 is this then let me the write what is TAN hyperbolic gamma? Tan hyperbolic gamma is nothing but TAN hyperbolic alpha + J beta.

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The image shows a handwritten derivation on a blue background. At the top, it states: $\tanh \gamma = \tanh(\alpha + j\beta) = \frac{\sinh \alpha \cosh \beta + j \cosh \alpha \sinh \beta}{\cosh \alpha \cosh \beta + j \sinh \alpha \sinh \beta}$. This is followed by a simplified form: $= \frac{\cosh \alpha \sinh \alpha + j \sinh \beta \cosh \beta}{\cosh^2 \alpha \cosh^2 \beta + \sinh^2 \alpha \sinh^2 \beta}$. Below this, two conditions are discussed. The first is: "when $\tanh \gamma$ is pure imaginary, $\cosh \alpha \sinh \alpha = 0$. \Downarrow $\sinh \alpha = 0 \Rightarrow \alpha = 0$." The second is: "when $\tanh \gamma$ is pure real, $\sinh \beta \cosh \beta = 0$ and $\cosh \alpha \sinh \alpha \neq 0$. \Downarrow $\sinh \alpha \neq 0$. $\alpha \neq 0$."

And that is equal to SIN hyperbolic alpha, COS beta + J COS hyperbolic alpha, SIN beta by COS hyperbolic alpha, COS beta + J SIN hyperbolic alpha, SIN beta. Now numerator denominator both are complex number A + JB so I generalized the determinant. So that becomes after some simple manipulation COS hyperbolic alpha COS hyperbolic alpha SIN hyperbolic + J SIN beta, COS beta by COS hyperbolic alpha square COS square beta + SIN hyperbolic square alpha and SIN square beta. So denominator is pure real now can talk of that.

When TAN hyperbolic gamma is we have seen in this case TAN hyperbolic is imaginary. So TAN hyperbolic is pure imaginary COS hyperbolic alpha, SIN hyperbolic alpha. So TAN hyperbolic alpha to be in a imaginary I require this part should go to 0 now here you see COS hyperbolic alpha always is greater than 0 so this is e to the power alpha + e to the power minus alpha by e to the power by 2 so it cannot be 0 COS hyperbolic alpha cannot be 0.

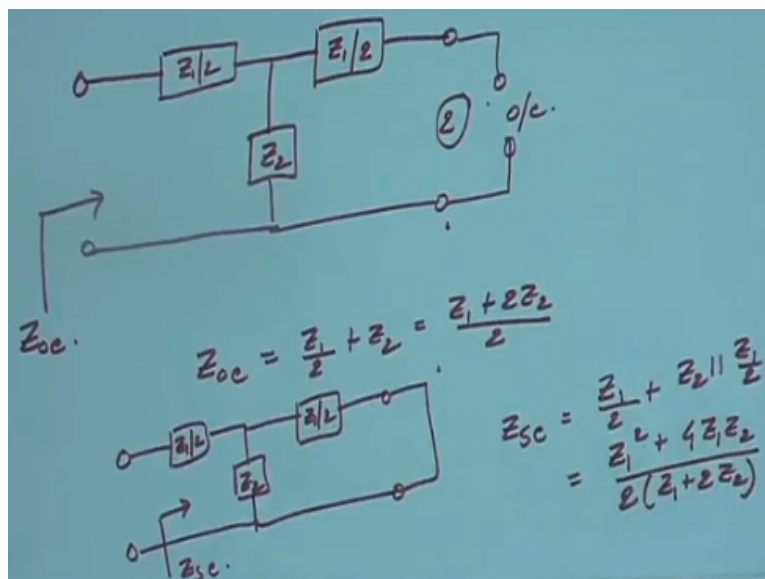
So the only possibilities this implies SIN hyperbolic alpha is equal to 0. What is SIN hyperbolic alpha that means e to the power alpha is equal to e to the power minus alpha that alpha is equal to 0. So where is the theorem, said that when Z0 is pure resistance alpha is equal to 0. You see Z0 is pure resistance from this I got alpha is equal to 0 so, okay let us see the other option. When

TAN hyperbolic gamma is pure real what I get you see this thing is pure real that means this part should go to 0.

I demand SIN beta, COS beta should be zero and also COS hyperbolic alpha, SIN hyperbolic alpha non-zero because this this also may go to zero then whole thing will be zero. But I require that this is nonzero but COS as I said COS is always greater than one so no problem it demands as COS this is always greater that equal to one. So what we demand is SIN hyperbolic alpha is nonzero so that means alpha nonzero means the attenuation constant is nonzero that mean there is a attenuation.

So that proves the theorem that when so what it shows let me go to my cases that it this case when Z_0 is real there will be no attenuation and when Z_0 is these there will be definitely attenuation. So can I not say that if I enforce this condition this is my pass band condition and this is my stop band condition of a filter. So I will do that, so but you also observe here that so this is my filter pass band I will go from here but here I have one demand that ZOC and ZSC they should be of opposites sign.

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Now ZOC ZSC they are that means that or I will see that, so let me now see my T section I can also see it with PIE section but the so it's a symmetrical and instead of earlier I was calling Z_1 now people call it $Z_1/2$ because there are 2 in series so okay. Now remember obviously Z_1

and Z_2 they are pure reactances we want to derive the propagation constant of this gamma of this network already we have done TAN gamma so from there we will try to see enforce the pass band condition.

So port 2 let us make that this is open circuit condition so that time we know we call it ZOC and what will be that $Z_{OC} = Z_1 \parallel 2Z_2$ that is equal to $\frac{Z_1 + 2Z_2}{2}$. Now I put that port 2 is shorted this is $Z_1 \parallel 2Z_2$ and this is ZSC. So what is ZSC it is you can easily see $Z_1 \parallel 2Z_2$ so that if you do it will become $Z_1 \parallel 2Z_2$ and that is equal to $\frac{Z_1 + 2Z_2}{2}$ and $Z_1 \parallel 2Z_2$ so I got ZOC and ZSC.

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$$Z_0 = \sqrt{Z_{oc} Z_{sc}} = \sqrt{\frac{Z_1^2 + 4Z_1 Z_2}{2}}$$

$$\tanh \gamma = \sqrt{\frac{Z_{sc}}{Z_{oc}}} = \sqrt{\frac{Z_1^2 + 4Z_1 Z_2}{(Z_1 + 2Z_2)^2}}$$

Condition $\rightarrow \frac{Z_1^2 + 4Z_1 Z_2}{(Z_1 + 2Z_2)^2} < 0$

Case I $Z_{oc} = +jX_a$
 $Z_{sc} = -jX_b$

$$\frac{Z_1 + 2Z_2}{2} = jX_a$$

$$(Z_1 + 2Z_2)^2 = -4X_a^2 < 0$$

So I can easily now find what is the characteristic impedance Z_0 I know always this is true I have expression for them say. So if you put it becomes $Z_1^2 + 4Z_1 Z_2$ by 2 already we have in that table that this gamma becomes pure imaginary. So this is the stop band there ZOC and ZSC there should be opposite side so, because that only makes, you see ZOC and ZSC their opposite side makes the characteristics impedance real otherwise they are of same type this is imaginary.

And what happens to the other, so characteristic impedance we have find what is the other parameter of the 2 port symmetrical reciprocal network gamma. So what is TAN gamma we know that is ZOC by ZSC and if you do this that will become root over $Z_1^2 + 4Z_1 Z_2$ by

$Z_1 + 2Z_2$ square. Now this we want this TAN gamma should be imaginary in the pass band. So our demand that in the pass band $Z_1^2 + 4Z_1 Z_2$ by $Z_1 Z_2 + 2Z_2^2$ square should be less than 0.

Consider this case 1 in the table will now confine ourselves to this case 1 and 2. In case 1 ZOC is $+jX_A$ this is $-jX_B$ so let us put it in case 1 ZOC $+jX_A$ ZSC is $-jX_B$ so what happens to $2Z_1 + 2Z_2$ by 2 this $Z_1 + 2Z_2$ by 2 becomes jX_A and $Z_1 + 2Z_2$ whole square that becomes $-4X_A^2$ squared we know X_A is positive real number so $Z_1 + 2Z_2$ square that is less than 0. So Tan beta that means in pass band we know this TAN beta value is this is less than 0.

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Case I.
 $Z_{oc} = -jX_A$
 $\frac{Z_1 + 2Z_2}{2} = -jX_A$
 $(Z_1 + 2Z_2)^2 = -4X_A^2 < 0$
 tanh γ to be imaginary $\rightarrow (Z_1^2 + 4Z_1Z_2) > 0$

So this whole thing less than 0 these need to be positive to satisfy this so we can say that TAN gamma to be imaginary $Z_1^2 + 4Z_1 Z_2$ that needs to be greater than 0. So this is from case one that means first case from pass band we know this. Let us see case 2 there we have seen ZOC is $-jX_A$ this is case 2 ZOC is $-jX_A$ ZSC $+jX_B$.

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subcase a $Z_1 = jX_1, Z_2 = jX_2.$
 $-X_1^2 - 4X_1X_2 > 0.$
 $(X_1^2 + 4X_1X_2) < 0.$

subcase b $Z_1 = -jX_1, Z_2 = -jX_2.$
 $(X_1^2 + 4X_1X_2) < 0.$

subcase c $Z_1 = jX_1, Z_2 = -jX_2.$
 $4X_1X_2 > X_1^2.$
 $4X_2 > X_1.$
 $0 < \frac{X_1}{4X_2} < 1$

So with this what happens to $Z_1 + Z_2$ by 2 that becomes $-jX_A$, so what happens to the square of this $Z_1 Z_2$ square that is again $-4X_A$ square that is again less than 0 so same condition that here also TAN gamma to be imaginary we require $Z_1^2 + 4Z_1 Z_2$ greater than 0. So the condition in the pass band if we want to have pass band without any attenuation then condition is $Z_1^2 + 4Z_1 Z_2$ greater than 0. Now Z_1 and Z_2 they are reactive elements.

So again there will be various sub cases so I can say sub case 1 a sub case A because there will be various cases of $Z_1 Z_2$. We should remember that $Z_1 = jX_1$ let us chose this $Z_2 = jX_2$. Then we can write what happens to $Z_1^2 + 4Z_1 Z_2$ that is $-X^2 - 4X_1 X_2$ greater than 0 or $X_1^2 + 4X_1 X_2$ is less than 0. So since X_1 and X_2 are positive real numbers this is not possible so I say that sub case is not possible similarly subcase B there will be $Z_1 = -jX_1$ is equal to $-jX_2$ again we will get the same condition $X_1^2 + 4X_1 X_2$ is less than 0.

So again not possible sub case C here let me take here as $Z_1 = jX_1, Z_2 = -jX_2$ if you do this what is this value $Z_1^2 + 4Z_1 Z_2$ always we are evaluating this $Z_1^2 + 4Z_1 Z_2$ that should be greater than 0. This is the condition to achieve the pass band. So we are doing this and getting this $4X_1 X_2$ is greater than X_1^2 now as X_1 is greater than 0. So $4X_2$ will be greater than X_1 or X_1 by $4X_2$ greater than 1, also X_1 is greater than 0.

(Refer Slide Time: 35:59)

$$\begin{aligned}
 & 0 < \frac{jX_1}{4X_2} < 1. \\
 & 0 > \frac{jX_1}{4(-jX_2)} > -1 \\
 & 0 > \frac{Z_1}{4Z_2} > -1
 \end{aligned}$$

subcase

$$Z_1 = -jX_1, Z_2 = +jX_2.$$

$$0 > \frac{Z_1}{4Z_2} > -1.$$

So combining these the condition is that 0 less than X_1 by $4X_2$ less than one or I can say 0 less than jX_1 by $4X_2$ is less than 1 or greater than jX_1 by $4 - jX_2$ is greater than -1 or 0 greater than Z_1 by $4Z_2$ is greater than -1. So this is possible that means here if we and enforce this condition it boils down to Z_1 by $4Z_2$ to this ratio should be between 0 and -1 and let us see the sub case D.

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$$\text{passband of filter} \rightarrow -1 < \frac{Z_1}{4Z_2} < 0$$

$$\text{stopband of filter} \rightarrow \begin{aligned} & a) \frac{Z_1}{4Z_2} > 0. \\ & b) \frac{Z_1}{4Z_2} < -1 \end{aligned}$$

That will be $Z_1 = -jX_1$ $Z_2 = +jX_2$ again if you do you do you get the condition that 0 Z_1 by $4Z_2$ $n - 1$. So this shows that, you can have the pass band of filter should be in the range this Z_1 one by four Z_2 this number should within this range. And where is the stop band of the filter,

there are two possibilities for that we have already seen that $Z_1 > 4Z_2$ that is greater than 0 or $Z_1 < 4Z_2$ that is less than 1.

This is mathematics we will see in the next lecture what is the implication of that but what we have done here, we have found out that if you want to have pass band the two reactances their impedances is Z_1 and Z_2 this $Z_1 > 4Z_2$ should be within this range then only you can have pass band and if you choose $Z_1 < 4Z_2$ either this or this you get stop band of filter Thank You